# **Phasors and Complex Numbers**

# Introduction

If you have a DC signal feeding a circuit, a single number can be used to describe the circuit. For example, consider the following circuit:



Voltage Divider: Y = 0.6X

From before, the output by voltage division is

$$Y = \left(\frac{R_1}{R_1 + R_2}\right) X$$
$$Y = \left(\frac{600}{600 + 400}\right) X$$
$$Y = 0.6 \cdot X$$

A good way to describe this circuit is to say it has a gain of 0.6: the output is 0.6 of the input.

If you change the circuit to be a capacitor and X(t) to be a sinusoid, however, the input / output relationship becomes a little more complicated



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If the input is a 628 rad/sec sine wave, it is also a 100Hz sine wave

$$\omega = 2\pi f$$
  
$$f = \frac{\omega}{2\pi} = \frac{628}{2\pi} = 100 Hz$$

The period is 10ms

$$T = \frac{1}{f} = \frac{1}{100Hz} = 0.01 \text{ sec}$$

If the input is sine wave, starting at t=0, with an amplitude of 100V, then

 $x(t) = 100\sin\left(628t\right)$ 

The output, however, is different than the input by both amplitude as well as a phase shift.

 $y(t) = 37\sin(628t - 67^0)$ 

You likewise need two numbers to describe this circuit

- An amplitude which tells you how much the signal is amplified, and
- An angle which tells you the phase shift of this circuit.

Complex numbers are a way to determine both of these with a single number.

## **Complex Numbers**

Let

 $j = \sqrt{-1}$ 

Any given number can then have a real and a complex part

x = a + jb

You can express this number in rectangular form (a + jb) or polar form

 $x = c \angle \theta$ 



Complex Number (a + jb) can also be expressed as  $(c \angle \theta)$ 

When you add complex numbers, the rectangular form is more convenient: the real part adds and the complex part adds.

$$(a_1+jb_1)+(a_2+jb_2)=(a_1+a_2)+j(b_1+b_2)$$

When you multiply complex numbers, polar form is more convenient: the amplitudes multiply, the angles add:

$$(c_1 \angle \theta_1)(c_2 \angle \theta_2) = c_1 c_2 \angle (\theta_1 + \theta_2)$$

Complex numbers let you express sine and cosine with a single number as well. The complex exponential is

 $e^{j\omega t} = \cos(\omega t) + j\sin(\omega t)$ 

If you multiply by a complex number

$$(a+jb) \cdot e^{j\omega t} = (a+jb) \cdot (\cos(\omega t) + j\sin(\omega t))$$
$$= (a\cos(\omega t) - b\sin(\omega t)) + j(-)$$

Taking the real part you get

 $a + jb \Rightarrow a\cos(\omega t) - b\sin(\omega t)$ 

The complex number a + jb represents a sine wave:

- The real part is the cosine term
- The complex part is the -sine term

With complex exponential, you can represent both sine and cosine with a single (complex) number.

#### **Complex Impedance's**

The impedance is the relationship between current and voltage. For a resistor

V = IR

The impedance of a resistor is R.

The impedance of a capacitor and inductor is a little trickier. First, assume all functions are of the form

 $x(t) = a \cdot e^{j\omega t}$ 

where 'a' could be complex. The VI characteristics for a capacitor is

 $I = C \frac{dV}{dt}$ 

If

then

$$\frac{dV}{dt} = j\boldsymbol{\omega} \cdot v_0 e^{j\boldsymbol{\omega} t}$$
$$\frac{dV}{dt} = j\boldsymbol{\omega} \cdot V$$

 $V(t) = v_0 e^{j\omega t}$ 

Then

$$I = C \cdot j\omega \cdot V$$
$$V = \left(\frac{1}{j\omega C}\right)I$$

The complex impedance of a capacitor is  $\left(\frac{1}{j\omega C}\right)$ 

The impedance of an inductor is from

$$V = L \frac{dI}{dt}$$

Assuming current is of the form

$$I = i_0 \cdot e^{j\omega t}$$

then

$$V = L \cdot j \omega \cdot I$$

The complex impedance of an inductor is  $j\omega L$ 

This lets you analyze any RLC circuit the same way you analyzed a resistor circuit - only the impedance's will be complex numbers.

## Example 1: RC Circuit

Determine y(t) for the following circuit:



Step 1: Replace the capacitor with its complex impedance. Since the input is 628 rad/sec, that's the frequency you care about

 $\omega = 628$  rad/sec  $Z_c = \frac{1}{j\omega c} = -j159\Omega$ 

Step 2: Solve just like you did with a DC circuit, only with complex numbers

$$Y = \left(\frac{-j159}{-j159+400}\right) X$$
  

$$Y = (0.37 \angle -68^{0}) \cdot 100 \sin(628t)$$
  

$$y(t) = 37 \sin(100t - 68^{0})$$

### In MATLAB:

```
>> C = 10e-6;
>> w = 628;
>> Zc = 1 / (j*w*C)
0 -159.24i
>> Gain = Zc / (400 + Zc)
0.1368 - 0.3436i
>> X = 0 - j*100; // x(t) = 100 sin(628t)
>> Y = Gain * X
-34.3632 -13.6796i
```

meaning

 $y(t) = -34.3632\cos(628t) + 13.6796\sin(628t)$ 

which is the rectangular form of y(t).

If you prefer polar form:

 $y(t) = 36.986 \cos(628t - 158^{\circ})$ 

In PartSim (www.PartSim.com)

Input the circuit using drag and drop



Make the input a sine wave with

- no DC offset
- 100V amplitude
- 100Hz

Enable DC Voltage					
DC Voltage	1.0				
Enable AC:					
AC Magnitude:	1.0				
AC Phase:	0				
Transient Source:					
None Sine Pulse	Exponential FM AM Noise				
Offset:	0				
Amplitude:	100				
Frequency:	100Hz				
Delay:	0.05				
Damping Factor:	0.0				

Run a transient response for

- 30ms (3 cycles)
- 10us step size

Run Simulation				
– Enable Simulations				
🔍 DC Bias 👘 DC Swee	p 🔲 AC Analysis 📝 Transient Response			
Configuration				
DC Sweep AC Response Tra	ansient Response			
Use Initial Conditions of Co	omponents			
Start Time:	Stop Time:			
0	30ms			
Time Step:	Max Step Size:			
10us	10us			
	Cancel	Run		

This results in a simulated input and output waveform:



Input (blue) and Output (black)

Note from the PartSim plot, the output after a short transient is

- 39Vp (as calculated)
- Delayed by 68 degrees from the input (as calculated)

To calculate the delay, use the following procedure:

- One cycle is 360 degrees or four divisions
- The output has zero crossings delayed by 0.8 divisions.

The phase shift is thus

$$\left(\frac{0.8 \text{div}}{4 \text{ div}}\right) 360^{\circ} = 72^{\circ}$$
 (actually 68 degrees but it's hard to read a graph that accurately)

Negative phase shift is a delay.

# Example 2: 3-Stage RC Circuit

Find the voltages for the following circuit when the input is

$$\mathbf{x}(t) = 100\,\cos(2t)$$



Step 1: Change the capacitors to their complex impedance (shown in red)

$$\omega = 2$$
  
0.01F:  $Z_c = \frac{1}{j\omega C} = -j50\Omega$   
0.02F:  $Z_c = \frac{1}{j\omega C} = -j25\Omega$   
0.03F:  $Z_c = \frac{1}{j\omega C} = -j16.67\Omega$ 

## Step 2: Write N equations for N unknowns

V1: 
$$\left(\frac{V_1-X}{100}\right) + \left(\frac{V_1}{150}\right) + \left(\frac{V_1}{-j50}\right) + \left(\frac{V_1-V_2}{200}\right) = 0$$

V2: 
$$\left(\frac{V_2 - V_1}{200}\right) + \left(\frac{V_2}{250}\right) + \left(\frac{V_2}{-j25}\right) + \left(\frac{V_2 - V_3}{300}\right) = 0$$

V3: 
$$\left(\frac{V_3-V_2}{300}\right) + \left(\frac{V_3}{350}\right) + \left(\frac{V_3}{-j16.67}\right) = 0$$

Step 3: Solve.

First, group terms

$$\left(\frac{1}{100} + \frac{1}{150} + \frac{1}{-j50} + \frac{1}{200}\right)V_1 + \left(\frac{-1}{200}\right)V_2 = \left(\frac{1}{100}\right)X$$
$$\left(\frac{-1}{200}\right)V_1 + \left(\frac{1}{200} + \frac{1}{250} + \frac{1}{-j25} + \frac{1}{300}\right)V_2 + \left(\frac{-1}{300}\right)V_3 = 0$$
$$\left(\frac{-1}{300}\right)V_2 + \left(\frac{1}{300} + \frac{1}{350} + \frac{1}{-j16.67}\right)V_3 = 0$$

Place in matrix form

$$\begin{bmatrix} \left(\frac{1}{100} + \frac{1}{150} + \frac{1}{-j50} + \frac{1}{200}\right) & \left(\frac{-1}{200}\right) & 0 \\ \left(\frac{-1}{200}\right) & \left(\frac{1}{200} + \frac{1}{250} + \frac{1}{-j25} + \frac{1}{300}\right) & \left(\frac{-1}{300}\right) \\ 0 & \left(\frac{-1}{300}\right) & \left(\frac{1}{300} + \frac{1}{350} + \frac{1}{-j16.67}\right) \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \\ V_3 \end{bmatrix} = \begin{bmatrix} \left(\frac{1}{100}\right) \\ 0 \\ 0 \end{bmatrix}$$

#### Put into MATLAB and solve

```
>> all = 1/100 + 1/150 - 1/(j*50) + 1/200;
>> a12 = -1/200;
>> a13 = 0;
>> a21 = -1/200;
>> a22 = 1/200 + 1/250 -1/(j*25) + 1/300;
>> a23 = -1/300;
>> a31 = 0;
>> a32 = -1/300;
>> a33 = 1/300+1/350-1/(j*16.67);
>> A = [a11,a12,a13;a21,a22,a23;a31,a32,a33]
   0.0217 + 0.0200i -0.0050
                                             0
           0.0123 + 0.0400i -0.0033
  -0.0050
                    -0.0033
                                        0.0062 + 0.0600i
       0
>> B = [1/100;0;0]
    0.0100
         0
         0
>> V = inv(A)*B*X
  24.2853 -23.2416i
  -1.7971 - 3.5726i
  -0.2066 + 0.0785i
```

meaning

$$V_1(t) = 24.28 \cos(2t) + 23.24 \sin(2t)$$
$$V_2(t) = -1.79 \cos(2t) + 3.57 \sin(2t)$$
$$V_1(t) = 0.21 \cos(2t) + 0.08 \sin(2t)$$

## If you prefer polar representation:

### meaning

$$V_1(t) = 33.61 \cos (2t - 43.7^{\circ})$$
$$V_2(t) = 3.999 \cos (2t - 116.7^{\circ})$$
$$V_3(t) = 0.22 \cos (2t + 159.2^{\circ})$$

# **PartSim Simulation**





Transient Response of Vin (blue), V1 (black), V2 (green), and V3 (orange)

Note that V1 has

- A peak of 33.686V (vs. 33.61V computed)
- A delay of  $\left(\frac{0.4 \text{ div}}{3.2 \text{ div}}\right) 360^{\circ} = 45^{\circ}$  (vs. 43.7 degrees computed)

Likewise, V2 and V3 match our calculations

	Vin	V1	V2	V3
Calculated	100V	33.61V	3.999V	220mV
PartSim	100V	33.678V	4.233V	592 mV