

ECE 461 - Homework #10

z-Transform, Converting G(s) to G(z). Due Monday, November 16th

1a) What is the difference equation the following transfer function represents?

$$Y = \left(\frac{0.01(z^2 + 1.3z + 0.4)}{z^3 - 3.2z^2 + 1.7z - 0.62} \right) X$$

Cross multiply

$$(z^3 - 3.2z^2 + 1.7z - 0.62)Y = (0.01(z^2 + 1.3z + 0.4))X$$

Note that zY means y(k+1)

$$y(k+3) - 3.2y(k+2) + 1.7y(k+1) - 0.62y(k) = 0.01(x(k+2) + 1.3x(k+1) + 0.4x(k))$$

or if you don't like future, do a change of variable

$$k+3 = k'$$

$$y(k') - 3.2y(k'-1) + 1.7y(k'-2) - 0.62y(k'-3) = 0.01(x(k'-1) + 1.3x(k'-2) + 0.4x(k'-3))$$

1b) Write a program to implement this filter.

```
while(1) {
    x3 = x2;
    x2 = x1;
    x1 = x0;
    x0 = A2D_Read();

    y3 = y2;
    y2 = y1;
    y1 = y0;
    y0 = 3.2*y1 - 1.7*y2 + 0.62*y3 + 0.01*(x1 + 1.3*x2 + 0.4*x3 );

    Wait();
}
```

2a) What is the difference equation the following transfer function represents?

$$Y = \left(\frac{0.03(z+1)(z+0.8)(z+0.6)}{z(z-0.9)(z-0.4)(z-0.2)} \right) X$$

Multiply it out

$$Y = \left(\frac{0.03(z^3+2.4z^2+1.88z+0.48)}{z^4-1.5z^3+0.62z^2-0.072z} \right) X$$

Cross multiply

$$(z^4 - 1.5z^3 + 0.62z^2 - 0.072z)Y = (0.03(z^3 + 2.4z^2 + 1.88z + 0.48))X$$

Note that zY means $y(k+1)$

$$y(k+4) - 1.5y(k+3) + 0.62y(k-2) - 0.072y(k+1) = 0.03(x(k+3) + 2.4x(k+2) + 1.88x(k+1) + 0.48x(k))$$

If you don't like future values, you can do a change of variable

$$k+4 = k'$$

$$y(k') - 1.5y(k'-1) + 0.62y(k'-2) - 0.072y(k'-3) = 0.03(x(k'-1) + 2.4x(k'-2) + 1.88x(k'-3) + 0.48x(k'-4))$$

Either answer is valid

2b) Write a program to implement this filter.

```
while(1) {
    x4 = x3;
    x3 = x2;
    x2 = x1;
    x1 = x0;
    x0 = A2D_Read();

    y3 = y2;
    y2 = y1;
    y1 = y0
    y0 = 1.5*y1 - 0.62*y2 + .072*y3 +
        0.03*(x1 + 2.4*x2 + 1.88*x3 + 0.48*x4 );

    Wait();
}
```

Note

- The new filter just changes one line of code
- It doesn't really matter if the poles or zeros are complex. All you care about are the coefficients of the polynomial expansion.

Problem 3-6) Assume a sampling rate of 100ms.

- Determine a filter $G(z)$ which has approximately the same step response as $G(s)$
- Plot the step response of $G(z)$ and $G(s)$ in VisSim (or similar program) to check your answer.

$$3) \quad G(s) = \left(\frac{30}{(s+2)(s+5)} \right)$$

Convert the poles to the z-plane as $z = e^{sT}$

$$\rightarrow s3 = [-2, -5]$$

$$\begin{aligned} &- 2. \\ &- 5. \end{aligned}$$

$$\rightarrow z3 = \exp(s3 * T)$$

$$\begin{aligned} &0.8187308 \\ &0.6065307 \end{aligned}$$

so

$$G(z) \approx \left(\frac{k}{(z-0.8187)(z-0.6065)} \right)$$

To find the gain on top, match the gain somewhere, like $s = 0$ ($z = +1$)

$$\left(\frac{30}{(s+2)(s+5)} \right)_{s=0} = 3$$

$$\left(\frac{k}{(z-0.8187)(z-0.6065)} \right)_{z=1} = 3$$

$$k = 0.2139717$$

If you want to match the delay, pick a frequency, such as $s = j2$ (the corner frequency). Add zeros at $z=0$ to match the phase shift

$$\left(\frac{30}{(s+2)(s+5)} \right)_{s=j2} = 1.9696 \angle -66.80^\circ$$

$$\left(\frac{0.2139}{(z-0.8187)(z-0.6065)} \right)_{s=j2} = 1.9572 \angle -78.92^\circ$$

Adding zeros at $z=0$ helps with the phase. Each zero adds

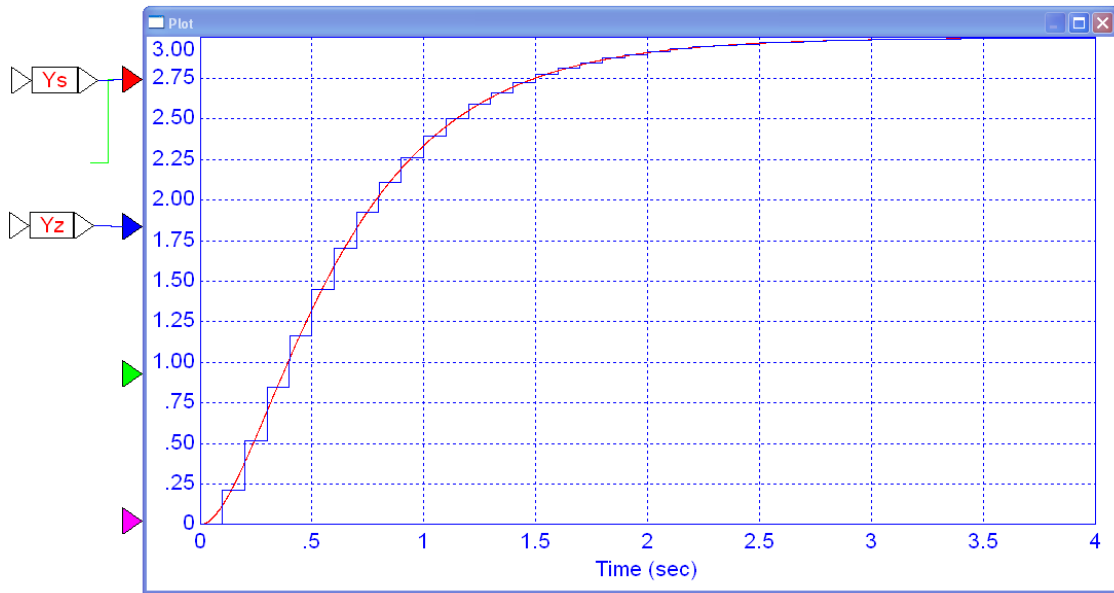
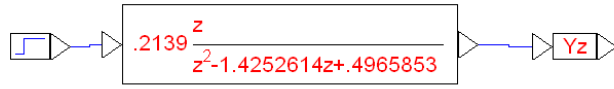
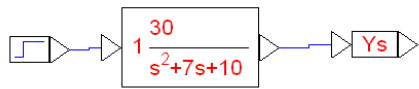
$$(e^{sT})_{s=j2} = 1 \angle +11.45^\circ$$

To fix the phase, you need to add 1.06 zeros

$$n = \frac{78.92^\circ - 66.8^\circ}{11.45^\circ} = 1.06$$

Rounding to 1.00 (add one zero at $z=0$) gives

$$G(z) \approx \left(\frac{0.2139z}{(z-0.8187)(z-0.6065)} \right)$$



$$4) \quad G(s) = \left(\frac{600}{(s^2+2s+15)(s+20)} \right)$$

Convert poles to the z plane as $z = e^{sT}$

```
-->s4 = [-1+j*3.74165, -1-j*3.74165, -20]'
```

```
- 1. - 3.74165i
- 1. + 3.74165i
- 20.
```

```
-->z4 = exp(s4*T)
```

```
0.8422346 - 0.3307139i
0.8422346 + 0.3307139i
0.1353353
```

```
-->poly(z4)
```

```
1. - 1.8198044 1.0466989 - 0.1108032
```

$$G(z) \approx \left(\frac{k}{(z-0.8422+j0.3307)(z-0.8422-j0.3307)(z-0.1353)} \right)$$

To find k, match the DC gain

$$\left(\frac{600}{(s^2+2s+15)(s+20)} \right)_{s=0} = 2$$

$$\left(\frac{k}{(z-0.8422+j0.3307)(z-0.8422-j0.3307)(z-0.1353)} \right)_{z=1} = 2$$

$$k = 0.2321826$$

This matches the DC gain. To find how many zeros belong, match the phase at some frequency, like $s = j1$

$$\left(\frac{600}{(s^2+2s+15)(s+20)} \right)_{s=j1} = 2.1186 \angle -10.99^\circ$$

$$\left(\frac{0.23218}{(z-0.8422+j0.3307)(z-0.8422-j0.3307)(z-0.1353)} \right)_{s=j1} = 0.2121 \angle -20.57^\circ$$

You can fix the phase by adding zeros at $z=0$. Each zero adds 5.73 degrees

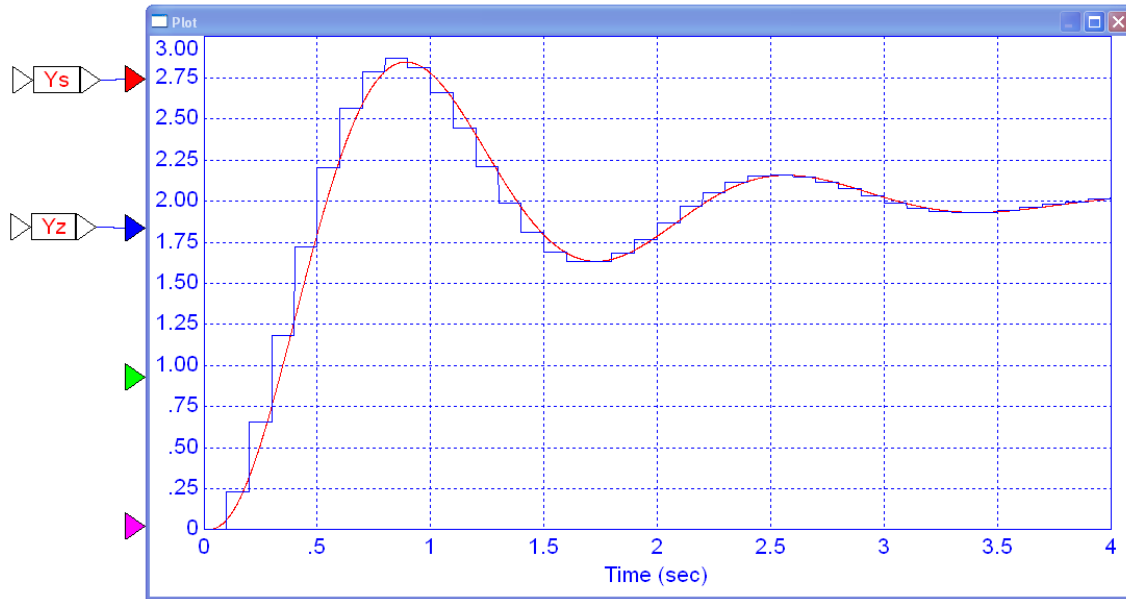
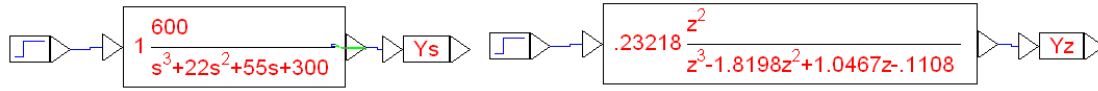
$$(e^{sT})_{s=j1} = 1 \angle 5.73^\circ$$

The number of zeros you need to make the phase match is 1.67

$$\left(\frac{20.57-10.99}{5.73} \right) = 1.67$$

Rounding to two zeros at $z=0$ results in

$$G(z) \approx \left(\frac{0.23218z^2}{(z-0.8422+j0.3307)(z-0.8422-j0.3307)(z-0.1353)} \right)$$



$$5) \quad G(s) = \left(\frac{625}{(s+1.31)(s+5.71)(s+12.45)(s+18.37)} \right) \quad (\text{heat equation from HW 5})$$

```
-->s5 = [-1.31, -5.71, -12.45, -19.37]'
```

```
s5 =
```

```
- 1.31
- 5.71
- 12.45
- 19.37
```

```
-->z5 = exp(s5*T)
```

```
z5 =
```

```
0.8772178
0.5649602
0.2879409
0.1441357
```

```
-->DC = 625 / prod(s5)
```

```
DC =
```

```
0.3464764
```

```
-->k = DC * prod(1 - z5)
```

```
k =
```

```
0.0112787
```

$$G(z) \approx \left(\frac{0.112787}{(z-0.8772)(z-0.5649)(z-0.2879)(z-0.1441)} \right)$$

To get a slightly better model, match the delay by matching the phase shift at $s = j1$

$$\left(\frac{625}{(s+1.31)(s+5.71)(s+12.45)(s+18.37)} \right)_{s=j1} = 0.2700 \angle -54.84^\circ$$

Add zeros at $z = 0$ to make the phase of $G(z)$ match at $s = j$

$$\left(\frac{0.112787}{(z-0.8772)(z-0.5649)(z-0.2879)(z-0.1441)} \right)_{s=j} = 0.2704 \angle -68.08^\circ$$

You can fix the phase by adding zeros at $z=0$. Each zero adds 5.7 degrees. The number of zeros you need is 2.31

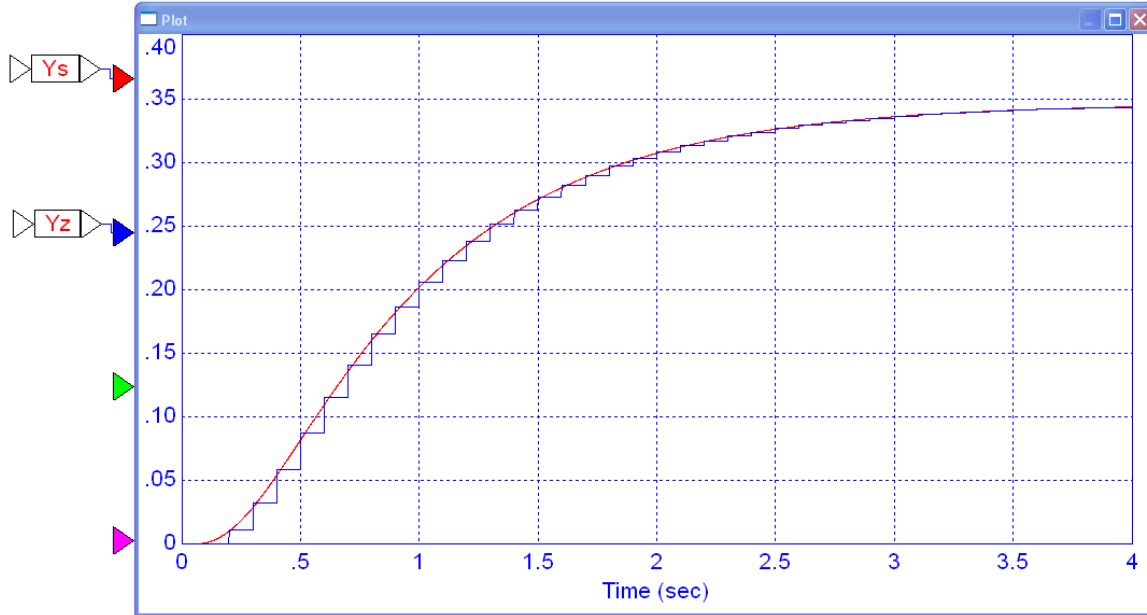
$$n = \left(\frac{68.08 - 54.84}{5.73} \right) = 2.31$$

Rounding to two zeros gives

$$G(z) \approx \left(\frac{0.112787z^2}{(z-0.8772)(z-0.5649)(z-0.2879)(z-0.1441)} \right)$$

Block 1:
$$1 \frac{625}{s^4 + 38.84s^3 + 472.013s^2 + 1930.9354s + 1803.8747}$$
 Output: Y_s

Block 2:
$$.0112787 \frac{z^2}{z^4 - 1.8742546z^3 + 1.1602271z^2 - .2739883z + .0205684}$$
 Output: Y_z



6) $G(s) = \left(\frac{2.5(s+0.25 \pm j3.86)}{(s \pm j4.319)(s \pm j1.157)(s+0.25 \pm j3.86)} \right)$ (3-mass system from HW 5)

First, convert the poles and zeros to the z-domain. The poles convert as:

```
-->s6 = [j*4.319, -j*4.319, j*1.157, -j*1.157, -0.25+j*3.86, -0.25-j*3.86]'
- 4.319i
  4.319i
- 1.157i
  1.157i
- 0.25 - 3.86i
- 0.25 + 3.86i

-->z6 = exp(s6*T)
0.9081721 - 0.4185971i
0.9081721 + 0.4185971i
0.9933142 - 0.1154420i
0.9933142 + 0.1154420i
0.9035490 - 0.3671903i
0.9035490 + 0.3671903i
```

The numerator (zeros) convert as:

```
-->s6n = [-0.25+j*3.86, -0.25-j*3.86]'
- 0.25 - 3.86i
- 0.25 + 3.86i

-->z6n = exp(s6n*T)
0.9035490 - 0.3671903i
0.9035490 + 0.3671903i
```

Matching the DC gain

```
-->DC = 2.5*prod(0 - s6n) / prod(0 - s6)
0.1001168

-->k = DC*prod(1 - z6) / prod(1 - z6n)
0.0002459
```

so G(z) is then:

$$G(z) \approx \left(\frac{0.0002459(z-0.9035 \pm j0.3671)}{(z-0.9081 \pm j0.4185)(z-0.9933 \pm j0.1154)(z-0.9035 \pm j0.3671)} \right)$$

Add two zeros at z=0 to match the phase (delay) in the step response

In VisSim:

note: you get numerical problems if you multiply out the polynomials. It works better numerically if you cascade second-order systems.

