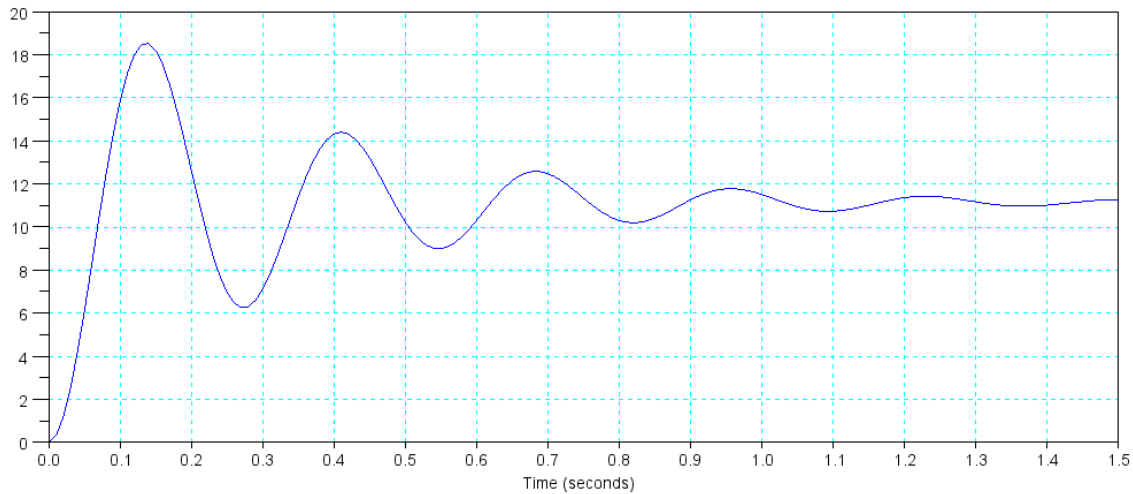


ECE 461 - Final: Name _____

Fall - 2020

Problem #1: 2nd Order Approximations

1a) Give the transfer function for a system with the following step response:



DC gain = 11.2

$$\text{Frequency of oscillation} = \left(\frac{3 \text{ cycles}}{0.82s} \right) 2\pi = 22.98 \frac{\text{rad}}{\text{sec}}$$

2% settling time = 1.3 seconds

$$\sigma = \frac{4}{1.3} = 3.07$$

$$G(s) \approx \left(\frac{6,020}{(s+3.07+j22.98)(s+3.07-j22.98)} \right)$$

1b) What is the step response for the following system:

$$Y = \left(\frac{2000}{(s+7+j12)(s+7-j12)(s+40)} \right) X$$

DC Gain	2% Settling Time	% Overshoot
0.2591	4/7 sec	16.0%

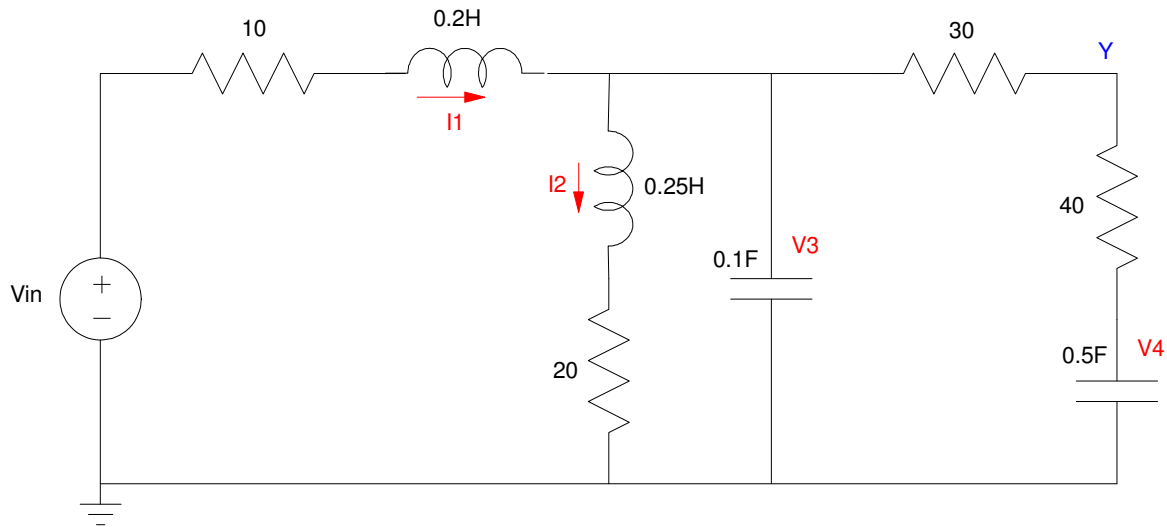
$$\theta = \arctan \left(\frac{12}{7} \right) = 59.74^\circ$$

$$\zeta = \cos \theta = 0.5039$$

$$OS = \exp \left(\frac{-\pi \zeta}{\sqrt{1-\zeta^2}} \right) = 16.0\%$$

Problem 2: Modeling and State Space

2a) Write the differential equations which describe the following circuit (i.e. write the N differential equations which correspond to the voltage node equations)



$$V_1 = 0.2\dot{i}_1 = V_{in} - 10I_1 - V_3$$

$$V_2 = 0.25\dot{i}_2 = V_3 - 20I_2$$

$$I_3 = 0.1\dot{V}_3 = I_1 - I_2 - \left(\frac{V_3 - V_4}{70}\right)$$

$$I_4 = 0.5\dot{V}_4 = \left(\frac{V_3 - V_4}{70}\right)$$

Output

$$\left(\frac{Y - V_3}{30}\right) + \left(\frac{Y - V_4}{40}\right) = 0$$

$$Y = \left(\frac{4}{7}\right)V_3 + \left(\frac{3}{7}\right)V_4$$

2b) Express these dynamics in state-space form

Group terms

$$\dot{i}_1 = 5V_{in} - 50I_1 - 5V_3$$

$$\dot{i}_2 = 4V_3 - 80I_2$$

$$\dot{V}_3 = 10I_1 - 10I_2 - \left(\frac{1}{7}\right)V_3 + \left(\frac{1}{7}\right)V_4$$

$$\dot{V}_4 = \left(\frac{1}{35}\right)V_3 - \left(\frac{1}{35}\right)V_4$$

Place in matrix form

$$s \begin{bmatrix} I_1 \\ I_2 \\ V_3 \\ V_4 \end{bmatrix} = \begin{bmatrix} -50 & 0 & -5 & 0 \\ 0 & -80 & 4 & 0 \\ 10 & -10 & \left(\frac{-1}{7}\right) & \left(\frac{1}{7}\right) \\ 0 & 0 & \left(\frac{1}{35}\right) & \left(\frac{-1}{35}\right) \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ V_3 \\ V_4 \end{bmatrix} + \begin{bmatrix} 5 \\ 0 \\ 0 \\ 0 \end{bmatrix} V_{in}$$

$$Y = \begin{bmatrix} 0 & 0 & \left(\frac{4}{7}\right) & \left(\frac{3}{7}\right) \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \\ V_3 \\ V_4 \end{bmatrix}$$

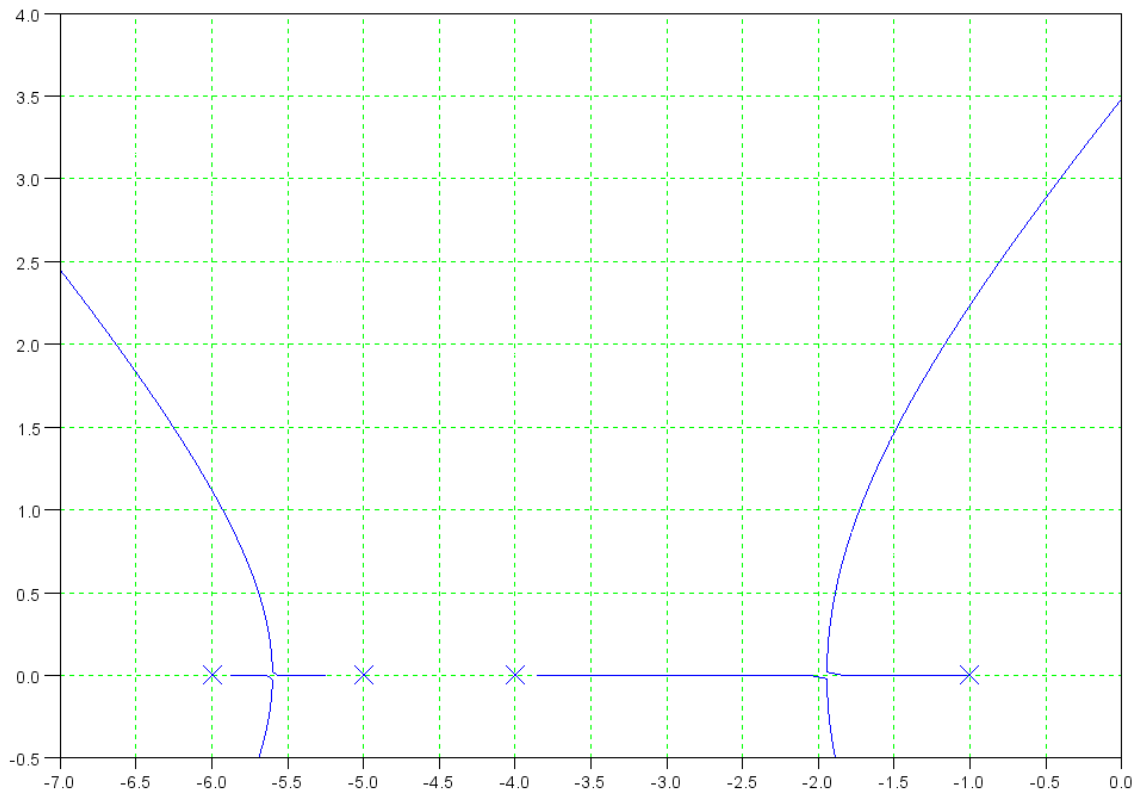
Problem 3: Root Locus

3) Gain Compensation: The root locus for

$$G(s) = \left(\frac{40}{(s+1)(s+4)(s+5)(s+6)} \right)$$

is shown below. Determine the following:

Maximum gain, k, for a stable closed-loop system	k = 20.3027 $s = j3.4821$
k for a damping ratio of 0.6	k = 2.9290
Closed-loop dominant pole(s) for a damping ratio of 0.6	s = -1.3224 + j1.7633
Closed-Loop DC gain for a damping ratio of 0.6	DC Gain = 0.4940 $K_p = (2.9290)(0.3333) = 0.9763$ $K_p / (1 + K_p) = 0.4940$



Problem #4: Compensator Design using Root Locus

4) Given the following stable system

$$G(s) = \left(\frac{100}{(s+0.5)(s+2)(s+6)} \right)$$

Determine a compensator, $K(s)$, which results in the closed-loop system having

- No error for a step input, and
- A closed-loop dominant pole at $s = -2 + j5$

Let

$$K(s) = k \left(\frac{(s+0.5)(s+2)}{s(s+a)} \right)$$

$$GK = \left(\frac{100k}{s(s+6)(s+a)} \right)$$

Evaluate what we know at $s = -2 + j5$

$$\left(\frac{100}{s(s+6)} \right)_{s=-2+j5} = 2.9001 \angle -163.1416^\circ$$

$$\angle(s+a) = 16.8584^\circ$$

$$a = 2 + \frac{5}{\tan(16.8584^\circ)} = 18.5$$

$$GK = \left(\frac{100k}{s(s+6)(s+18.5)} \right)$$

Evaluate what we know

$$GK = \left(\frac{100}{s(s+6)(s+18.5)} \right)_{s=-2+j5} = 0.1682 \angle 180^\circ$$

$$k = \frac{1}{0.1682} = 5.945$$

so

$$K(s) = \left(\frac{5.945(s+0.5)(s+2)}{s(s+18.5)} \right)$$

Problem #5: Discrete-Time Compensator Design

5) Given the following stable system

$$G(s) = \left(\frac{100}{(s+0.5)(s+2)(s+6)} \right)$$

Determine a digital compensator, $K(z)$, which results in the closed-loop system having

- No error for a step input,
- A closed-loop dominant pole at $s = -2 + j5$, and
- A sampling rate of $T = 0.2$

Let

$$K(s) = k \left(\frac{(s+0.5)(s+2)}{s(s+a)} \right)$$

or in the z-plane

$$K(z) = k \left(\frac{(z-0.9048)(z-0.3679)}{(z-1)(z-a)} \right)$$

$G(s) * K(z) * \text{zero order hold}$ is (modeled as a 1/2 sample delay)

$$GK\Delta = \left(\frac{100}{(s+0.5)(s+2)(s+6)} \right) \cdot k \left(\frac{(z-0.9048)(z-0.3679)}{(z-1)(z-a)} \right) \cdot e^{-0.1s}$$

Evaluate what we know at

- $s = -2 + j5$
- $z = e^{sT} = 0.3622 + j0.5641$

$$\left(\frac{100}{(s+0.5)(s+2)(s+6)} \right)_{s=-2+j5} \cdot \left(\frac{(z-0.9048)(z-0.3679)}{(z-1)} \right)_{z=0.3622+j0.5641} \cdot (e^{-0.1s})_{s=-2+j5} = 0.8433 \angle 54^\circ$$

Too much phase. Try

$$K(s) = k \left(\frac{(s+0.5)(s+2)(s+6)}{s(s+a)^2} \right)$$

$$K(z) = k \left(\frac{(z-0.9048)(z-0.3679)(s-0.3012)}{(z-1)(z-a)^2} \right)$$

Now

$$(G(s) \cdot K(z) \cdot e^{-sT/2})_{s=-2+j5} = 0.2150 \angle -106.897^\circ$$

$$\angle(z-a)^2 = 73.1026^\circ$$

$$\angle(z-a) = 36.5513^\circ$$

$$a = 0.3622 - \frac{0.5641}{\tan(36.5513^\circ)} = -0.3987$$

and

$$K(z) = k \left(\frac{(z-0.9048)(z-0.3679)(s-0.3012)}{(z-1)(z+0.3977)^2} \right)$$

Evaluate what we know

$$(G(s) \cdot K(z) \cdot e^{-sT/2})_{s=-2+j5} = 0.2401 \angle 180^\circ$$

$$k = \frac{1}{0.2410} = 4.1655$$

and

$$K(z) = 4.1655 \left(\frac{(z-0.9048)(z-0.3679)(s-0.3012)}{(z-1)(z+0.3977)^2} \right)$$

Problem #6: Compensator Design using Bode Plots

6) Given the following stable system

$$G(s) = \left(\frac{100}{(s+0.5)(s+2)(s+6)} \right)$$

Determine a compensator, $K(s)$, which results in the closed-loop system having

- A closed-loop DC gain of 1.000 (i.e. no error for a step input),
- A 0dB gain frequency of 5 rad/sec, and
- $M_m = 1.45$

Convert M_m to a phase margin

$$\frac{1}{M_m^2} = 2 + 2 \cos \phi$$

$$\phi = -139.6575^\circ$$

Pick $K(s)$ so that

$$G(j5)K(j5) = 1 \angle -139.6575^\circ$$

Let

$$K(s) = k \left(\frac{(s+0.5)(s+2)}{s(s+a)} \right)$$

$$GK = \left(\frac{100k}{s(s+6)(s+a)} \right)$$

Evaluating what we know at 5 rad/sec

$$\left(\frac{100}{s(s+6)} \right)_{s=j5} = 0.2561 \angle -129.8056^\circ$$

To make the phase add up to -139.6575 degrees

$$\angle(s+a) = 9.8519^\circ$$

$$a = \frac{5}{\tan(9.8519^\circ)} = 28.7913$$

and

$$GK = \left(\frac{100k}{s(s+6)(s+28.7913)} \right)$$

Evaluate what we know

$$\left(\frac{100}{s(s+6)(s+28.7913)} \right)_{s=j5} = 0.0876 \angle -139.6575^\circ$$

meaning

$$k = \frac{1}{0.0876} = 11.4116$$

and

$$K(s) = 11.4116 \left(\frac{(s+0.5)(s+2)}{s(s+28.7913)} \right)$$

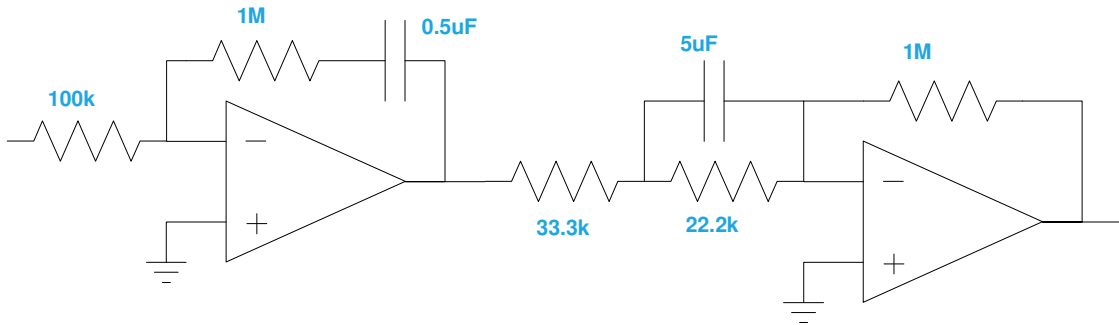
Problem #8: Implementation

7) Determine R and C so that the following compensator has the transfer function of

$$K(s) = 300 \left(\frac{(s+2)(s+9)}{s(s+15)} \right)$$

Rewrite as

$$K(s) = \left(\frac{10(s+2)}{s} \right) \left(30 \left(\frac{s+9}{s+15} \right) \right)$$



There are other solutions.