

# Homework #9: ECE 461/661

z-Transforms, s to z conversion, Root Locus in the z-Domain. Due Monday, November 9th

## z-Transforms

1) Determine the difference equation that relates X and Y

$$Y = \left( \frac{0.005z}{(z-0.95)(z-0.9)(z-0.5)} \right) X$$

Cross multiply

$$(z - 0.95)(z - 0.9)(z - 0.5)Y = (0.005z)X$$

-->poly([0.95, 0.9, 0.5])

ans = 1. - 2.35 1.78 - 0.4275

$$(z^3 - 2.35z^2 + 1.78z - 0.4275)Y = (0.005z)X$$

'zY' means 'the next value of y(k)' or 'y(k+1)'

$$y(k+3) - 2.35 y(k+2) + 1.78 y(k+1) - 0.4275 y(k) = 0.005 x(k+1)$$

or using a change in variable

$$y(k) - 2.35 y(k-1) + 1.78 y(k-2) - 0.4275 y(k-3) = 0.005 x(k-2)$$

Either answer is correct

2) Determine  $y(k)$  assuming

$$Y = \left( \frac{0.005z}{(z-0.95)(z-0.9)(z-0.5)} \right) X \quad x(k) = u(k)$$

Substitute the z-transform for  $x(k)$

$$Y = \left( \frac{0.005z}{(z-0.95)(z-0.9)(z-0.5)} \right) \left( \frac{z}{z-1} \right)$$

Pull out a  $z$  (we'll need this later)

$$Y = \left( \frac{0.005z}{(z-1)(z-0.95)(z-0.9)(z-0.5)} \right) z$$

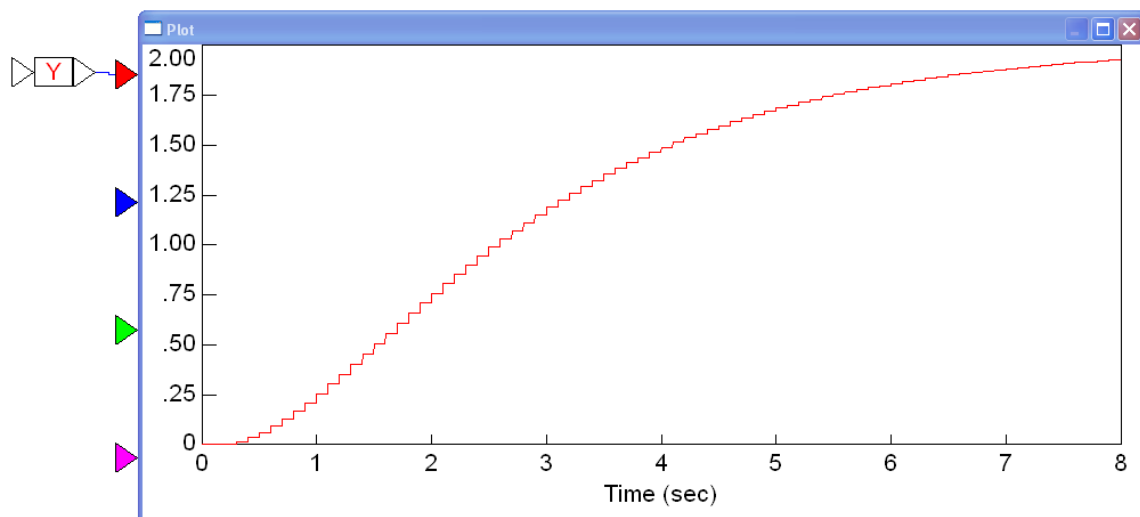
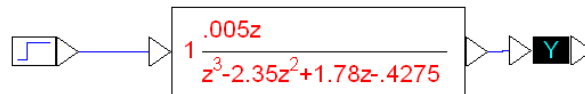
Do partial fractions

$$Y = \left( \left( \frac{2}{z-1} \right) + \left( \frac{-4.4444}{z-0.95} \right) + \left( \frac{2.25}{z-0.9} \right) + \left( \frac{-0.0278}{z-0.5} \right) \right) z$$

$$Y = \left( \frac{2z}{z-1} \right) + \left( \frac{-4.4444z}{z-0.95} \right) + \left( \frac{2.25z}{z-0.9} \right) + \left( \frac{-0.0278z}{z-0.5} \right)$$

Take the inverse z-transform

$$y(k) = \left( 2 - 4.4444(0.95)^k + 2.25(0.9)^k - 0.0278(0.5)^k \right) u(k)$$



3) Determine  $y(t)$  assuming

$$Y = \left( \frac{0.005z}{(z^2 - 1.6z + 0.68)(z - 0.5)} \right) X \quad x(k) = u(k)$$

Plug in the z-transform for  $x(k)$ . Factor the roots

$$Y = \left( \frac{0.005z}{(z - 0.8 + j0.2)(z - 0.8 - j0.2)(z - 0.5)} \right) \left( \frac{z}{z - 1} \right)$$

Pull out a  $z$

$$Y = \left( \frac{0.005z}{(z - 1)(z - 0.8 + j0.2)(z - 0.8 - j0.2)(z - 0.5)} \right) z$$

Do partial fractions

$$Y = \left( \left( \frac{0.125}{z - 1} \right) + \left( \frac{-0.0433 - j0.0913}{z - 0.8 + j0.2} \right) + \left( \frac{-0.0433 + j0.0913}{z - 0.8 - j0.2} \right) + \left( \frac{-0.0385}{z - 0.5} \right) \right) z$$

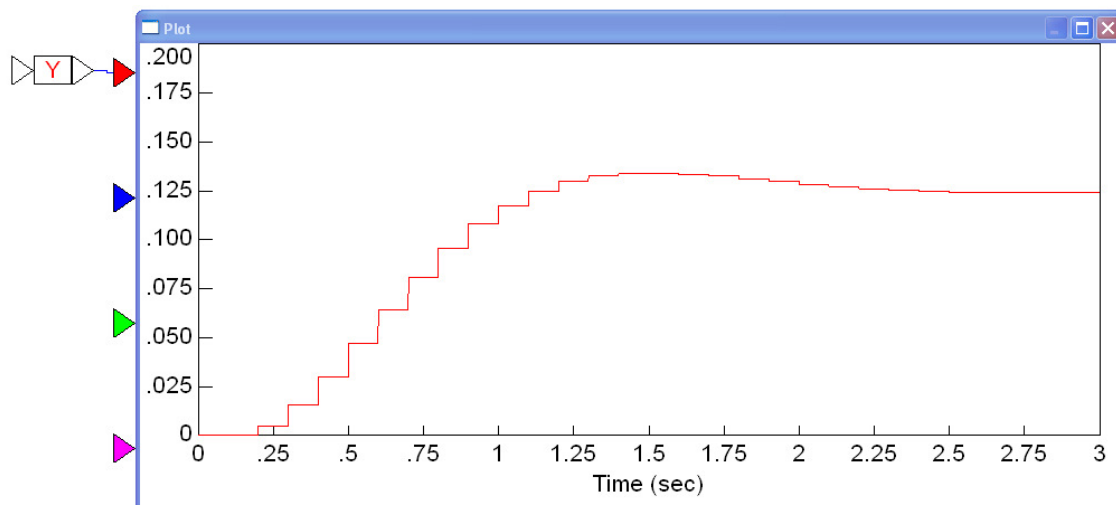
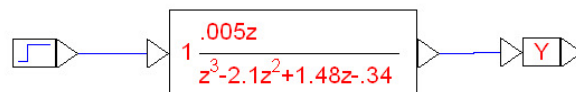
$$Y = \left( \frac{0.125z}{z - 1} \right) + \left( \frac{-0.0433 - j0.0913}{z - 0.8 + j0.2} \right) z + \left( \frac{-0.0433 + j0.0913}{z - 0.8 - j0.2} \right) z + \left( \frac{-0.0385z}{z - 0.5} \right)$$

polar form actually works better with z transforms

$$Y = \left( \frac{0.125z}{z - 1} \right) + \left( \frac{0.1011 \angle -115.3^\circ}{z - 0.8246 \angle -14.04^\circ} \right) z + \left( \frac{0.1011 \angle 115.3^\circ}{z - 0.8246 \angle 14.04^\circ} \right) z + \left( \frac{-0.0385z}{z - 0.5} \right)$$

Take the inverse z-transform

$$y(k) = \left( 0.125 + 0.2022(0.8246)^k \cos(14.04^\circ \cdot k + 115.3^\circ) - 0.0385(0.5)^k \right) u(k)$$



## s to z conversion

3) Determine the discrete-time equivalent of  $G(s)$ . Assume  $T = 0.5$  second

$$G(s) = \left( \frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right)$$

The conversion from the s to z plane is

$$z = e^{sT}$$

$$s = -0.1617 \quad z = 0.9223$$

$$s = -1.04 \quad z = 0.5945$$

$$s = -2.719 \quad z = 0.2568$$

$$s = -5.05 \quad z = 0.0801$$

giving

$$G(z) \approx \left( \frac{k}{(z-0.9223)(z-0.5945)(z-0.2568)(z-0.0801)} \right)$$

Matching the DC gain

$$\left( \frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right)_{s=0} = 0.6248$$

$$\left( \frac{k}{(z-0.9223)(z-0.5945)(z-0.2568)(z-0.0801)} \right)_{z=1} = 0.6248$$

$$k = 0.0135$$

$$G(z) \approx \left( \frac{0.0135}{(z-0.9223)(z-0.5945)(z-0.2568)(z-0.0801)} \right)$$

(optional)  $G(z)$  has too much delay. You can fix this by adding zeros at  $z=0$ .

To determine how many zeros you need, match the phase shift at  $s = j0.1$  rad/sec

$$\left( \frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right)_{s=j0.1} = 0.5285 \angle -40.4669^\circ$$

$$z = e^{sT} = e^{(j0.1)(0.5)} = e^{j0.05} = 1 \angle 0.05$$

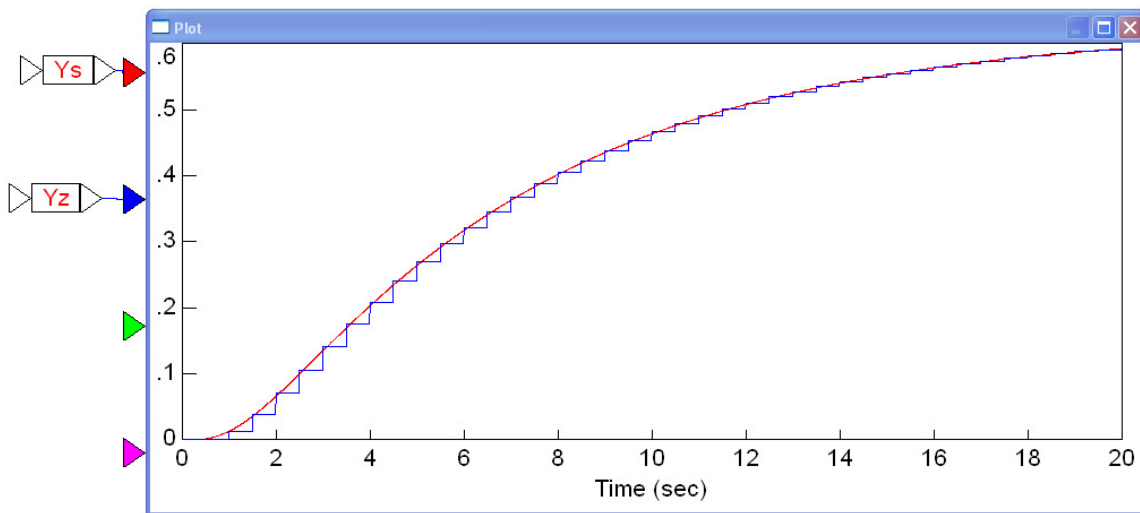
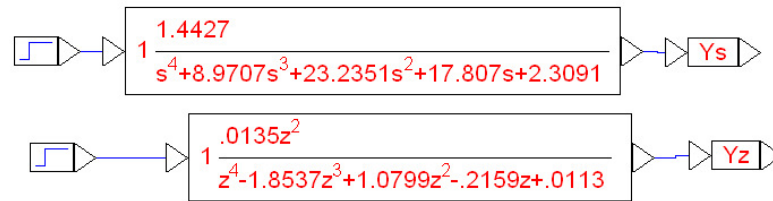
$$\left( \frac{0.0135}{(z-0.9223)(z-0.5945)(z-0.2568)(z-0.0801)} \right)_{z=1 \angle 0.05} = 0.5304 \angle -47.1903^\circ$$

for a difference of 6.7234 degrees. Each zero at  $z=0$  adds 0.05 radians = 2.86 degrees. The number of zeros you need to make the phase match is

$$n = \left( \frac{6.7234^0}{2.86^0} \right) = 2.34$$

Round to  $n = 2$

$$G(z) \approx \left( \frac{0.0135z^2}{(z-0.9223)(z-0.5945)(z-0.2568)(z-0.0801)} \right)$$



4) Determine the discrete-time equivalent of  $G(s)$ . Assume  $T = 0.1$  second

$$G(s) = \left( \frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right)$$

Converting from the s-plane to the z-plane

$$s = -0.1617 \quad z = 0.9840$$

$$s = -1.04 \quad z = 0.9012$$

$$s = -2.719 \quad z = 0.7619$$

$$s = -5.05 \quad z = 0.6035$$

giving

$$G(z) \approx \left( \frac{k}{(z-0.9840)(z-0.9012)(z-0.7619)(z-0.6035)} \right)$$

Pick  $k$  to match the DC gain

$$G(s = 0) = 0.6248$$

$$\left( \frac{k}{(z-0.9840)(z-0.9012)(z-0.7619)(z-0.6035)} \right)_{z=1} = 0.6248$$

$$k = 0.0000932$$

$$G(z) \approx \left( \frac{0.0000932}{(z-0.9840)(z-0.9012)(z-0.7619)(z-0.6035)} \right)$$

(optional) Add zeros at  $z=0$  to match the phase shift at  $s = j0.1$

$$z = e^{sT} = e^{(j0.1)(0.1)} = e^{j0.01} = 1 \angle 0.01$$

$$\left( \frac{0.0000932}{(z-0.9840)(z-0.9012)(z-0.7619)(z-0.6035)} \right)_{z=1 \angle 0.01} = 0.5279 \angle 41.718^\circ$$

$$\left( \frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right)_{s=j0.1} = 0.5285 \angle -40.4669^\circ$$

for a difference of 1.2513 degrees (0.0218 radian). Each zero at  $z=0$  adds 0.01 radian, meaning you need 2.18 zeros. Let  $n=2$

$$G(z) \approx \left( \frac{0.0000932z^2}{(z-0.9840)(z-0.9012)(z-0.7619)(z-0.6035)} \right)$$

**Note:** If you change the sampling rate,  $G(z)$  changes.

- Any compensator you designed for  $G(z)$  at  $T = 0.5$  won't work for  $T = 0.1$ .
- Changing the sampling rate is a big deal

## Root Locus in the z-Domain

5) Assume  $T = 0.5$  seconds.

$$G(s) = \left( \frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right)$$

Draw the root locus for  $G(z)$

From problem #3

$$G(z) \approx \left( \frac{0.0135z^2}{(z-0.9223)(z-0.5945)(z-0.2568)(z-0.0801)} \right)$$

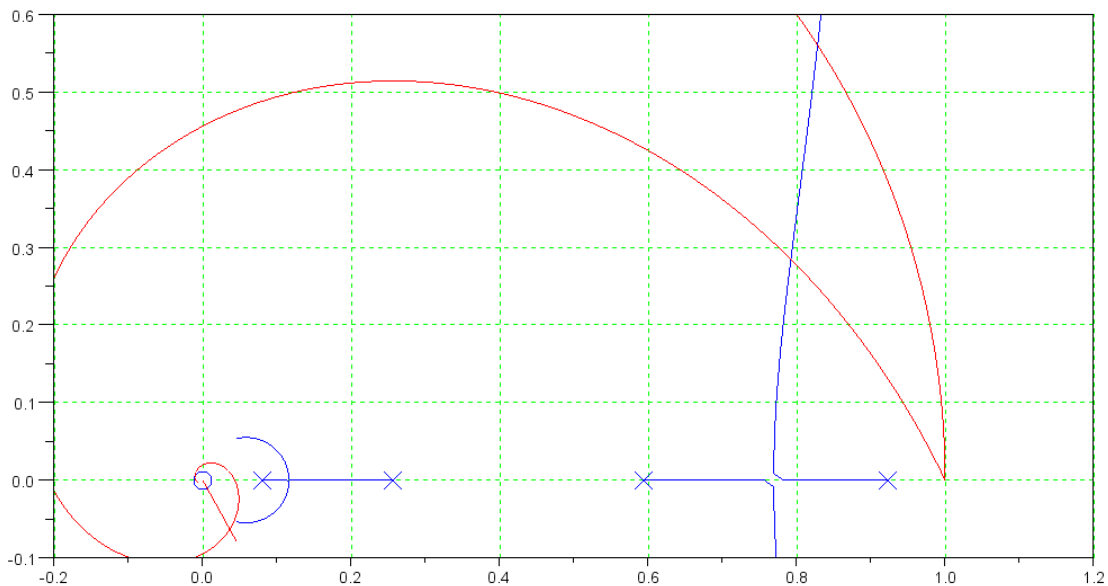
Drawing the root locus in Matlab

```
G = zpk([0,0],[0.9223,0.5945,0.2568,0.0801],0.0135);  
k = logspace(-2,2,1000)';
```

```
rlocus(G,k);  
hold on
```

```
s = [0:0.01:10] * (-1+j*2);  
T = 0.5;  
z = exp(s*T);  
plot(real(z), imag(z), 'r');
```

```
s = [0:0.01:10] * (0+j*1);  
z = exp(s*T);  
plot(real(z), imag(z), 'r');
```



6) Find k for no overshoot in the step response

- Simulate the step response of the closed-loop system

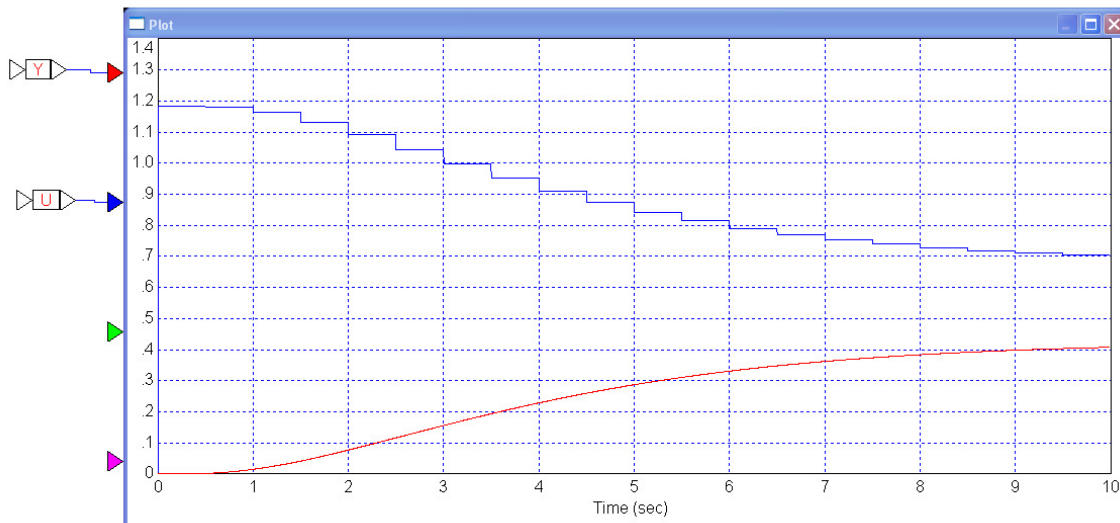
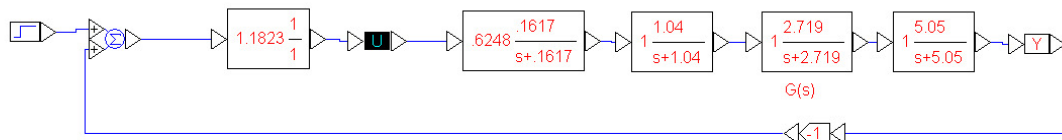
```
G = zpk([0,0],[0.9223,0.5945,0.2568,0.0801],0.0135);
```

```
z = 0.77
evalfr(G,z)
```

```
- 0.8457956
```

```
k = 1/abs(ans)
```

```
k = 1.1823188
```





7) Find k for a damping ratio of 0.4559 (20% overshoot)

- Simulate the step response of the closed-loop system

```
G = zpk([0,0],[0.9223,0.5945,0.2568,0.0801],0.0135);
```

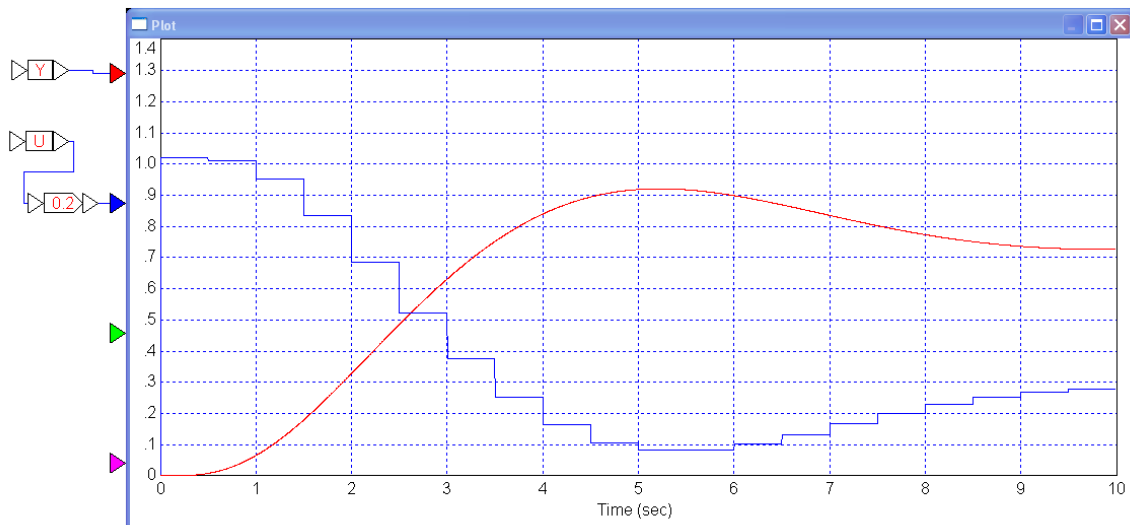
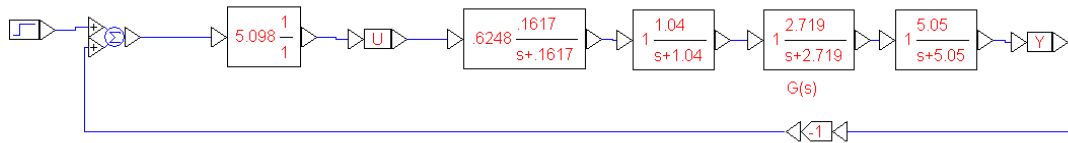
```
z = 0.78 + j*0.28;
```

```
evalfr(G,z)
```

```
- 0.1960269 + 0.0132974i
```

```
k = 1/abs(ans)
```

```
k = 5.0896433
```



### 8) Find k for a damping ratio of 0.000

- Simulate the step response of the closed-loop system

```
G = zpk([0,0],[0.9223,0.5945,0.2568,0.0801],0.0135);
```

```
z = 0.82 + j*0.56;  
evalfr(G,z)
```

```
- 0.0525318 + 0.0015737i
```

```
k = 1/abs(ans)
```

```
k = 19.027567
```

