

Homework #10: ECE 461/661

Digital PID Control. Due Monday, November 16th

I Control

Assume $T = 0.5$ seconds:

$$G(s) = \left(\frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right)$$

1) Design a digital I controller

$$K(s) = k \left(\frac{z}{z-1} \right)$$

that results in 20% overshoot in the step response.

Simulate the step response of the closed-loop system (VisSim or Simulink preferred with $K(z)*G(s)$)

Option #1: Design in the z-domain. Convert $G(s)$ to $G(z)$

```
T = 0.5;
s = [-0.1617, -1.04, -2.719, -5.05]
Gz = zpk([0,0],exp(s*T),1, T);
Gs = zpk([],s,1.4427);
k = evalfr(Gs,0) / evalfr(Gz,1)
    k = 0.0135
Gz = zpk([0,0],exp(s*T),k, T);
zpk(Gz)

          0.0135 z^2
-----
(z-0.9223) (z-0.5945) (z-0.2568) (z-0.08006)

Sampling time (seconds): 0.5
```

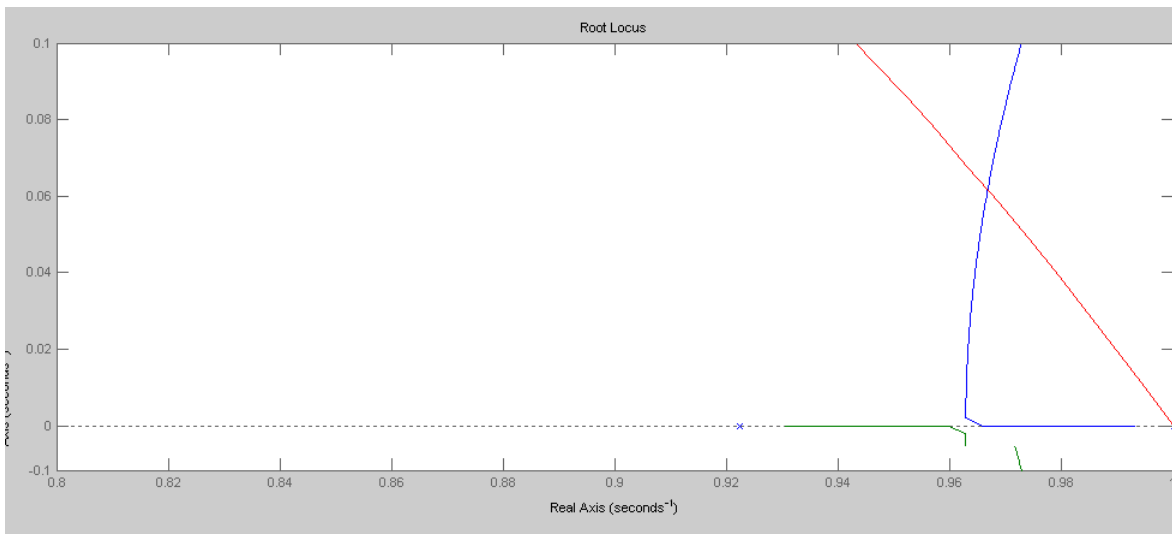
$$G(z) \approx \left(\frac{0.0135z^2}{(z-0.9223)(z-0.5945)(z-0.2568)(z-0.0801)} \right)$$

Add $K(z)$

$$GK = \left(\frac{0.0135z^3}{(z-1)(z-0.9223)(z-0.5945)(z-0.2568)(z-0.0801)} \right)$$

Draw the root locus along with the damping line

```
k = logspace(-2,2,1000)';
rlocus(GzKz,k);
hold on;
s = (-1 + j*2) * [0:0.01:10]';
z = exp(s*T);
plot(real(z),imag(z),'r');
```



Result: $z = 0.9667 + j0.0616$

Option #2: Numerical Solution.

The plant + compensator + sample & hold is

$$G \cdot K \cdot H = \left(\frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right) \cdot \left(\frac{kz}{z-1} \right) \cdot (e^{-sT/2})$$

Search along the damping line until the angles add up to 180 degrees

```

Kz = zpk(0, 1, 1, T);
T = 0.5;
s = -0.1 + j*0.2;
z = exp(s*T);
evalfr(Gs, s) * evalfr(Kz, z) * exp(-s*T/2)

-4.3508 + 2.3219i

s = 0.9*s;
z = exp(s*T);
evalfr(Gs, s) * evalfr(Kz, z) * exp(-s*T/2)

-5.4439 + 2.1500i

```

time passes....

```

s = 1.0001*s;
z = exp(s*T);
evalfr(Gs, s) * evalfr(Kz, z) * exp(-s*T/2)

-9.7937 - 0.0001i

s = -0.0630 + 0.1260i
z = 0.9671 + 0.0610i

```

K(z) is then

```
evalfr(Gs,s) * evalfr(Kz,z) * exp(-s*T/2)
```

```
ans = -9.7937 - 0.0001i
```

```
k = 1 / abs(ans)
```

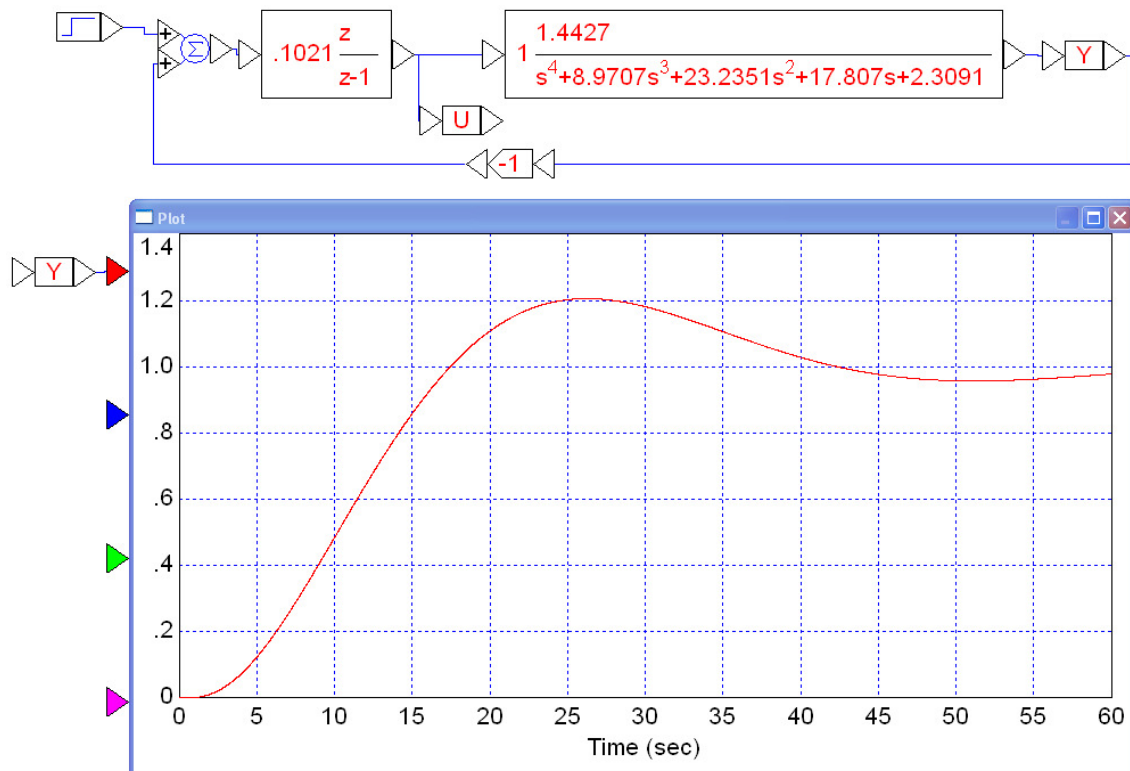
```
k = 0.1021
```

```
Kz = zpk(0,1,k, T)
```

$$\frac{0.1021 z}{(z-1)}$$

Sampling time (seconds): 0.5

Checking with VisSim



PI Control

2) Design a digital PI controller that results in 20% overshoot in the step response.

Use option #2 (numerical method). Choose the zero to cancel the pole at $s = -0.1617$

$$z = e^{sT} = 0.922$$

$$K(z) = k \left(\frac{z-0.922}{z-1} \right)$$

Search along the dampin line until the angles add up to 180 degrees

$$G(s) \cdot K(z) \cdot H(s) = 1 \angle 180^\circ$$

$$\left(\left(\frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right) \cdot \left(\frac{k(z-0.922)}{z-1} \right) \cdot (e^{-sT/2}) \right)_s = 1 \angle 180^\circ$$

itterating results in

$$s = -0.2776 + j0.5553$$

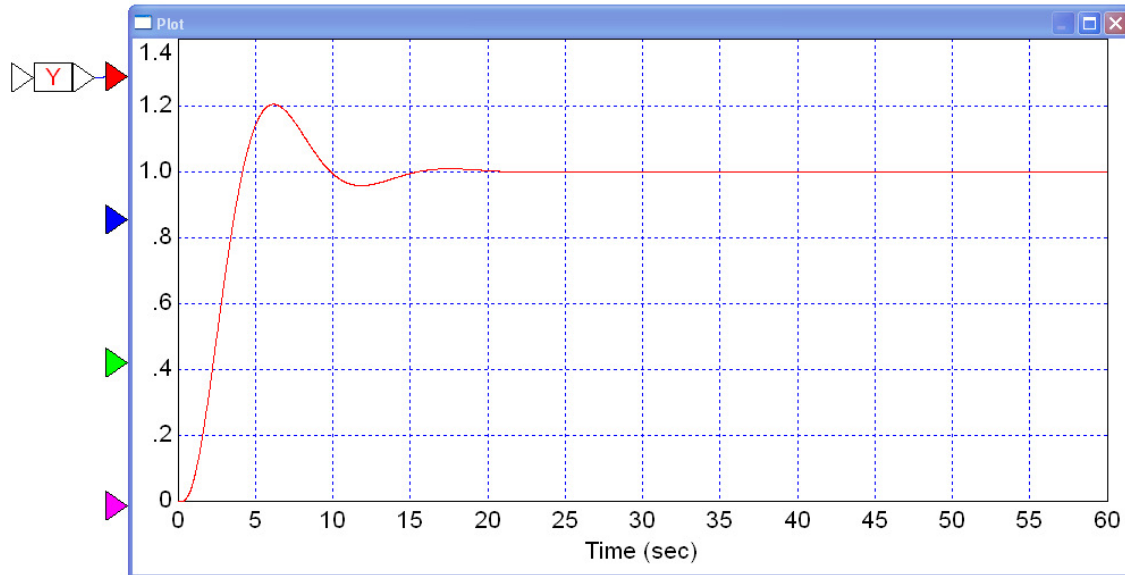
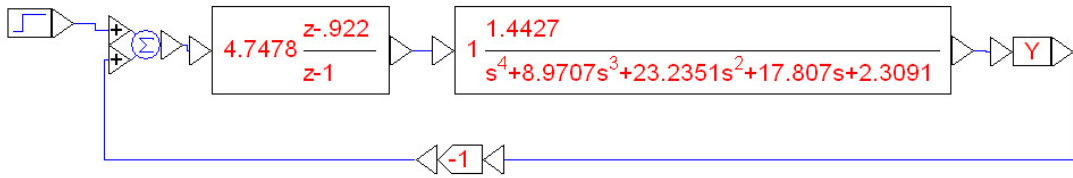
$$z = 0.8371 + j0.2386$$

$$\left(\left(\frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right) \cdot \left(\frac{k(z-0.922)}{z-1} \right) \cdot (e^{-sT/2}) \right)_s = 0.2016 \angle 180^\circ$$

$$k = \frac{1}{0.2016} = 4.7478$$

and

$$K(z) = 4.7478 \left(\frac{z-0.922}{z-1} \right)$$



PID Control

3) Design a digital PID controller that results in 20% overshoot in the step response.

Let $K(z)$ cancel the poles at $s = -0.1617$ and $s = -1.04$

$$K(z) = k \left(\frac{(z-0.922)(z-0.5945)}{z(z-1)} \right)$$

The plant + compensator + sample and hold is now

$$G(s) \cdot K(z) \cdot H(s) = 1 \angle 180^\circ$$

$$\left(\left(\frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right) \cdot \left(\frac{k(z-0.922)(z-0.5945)}{z(z-1)} \right) \cdot (e^{-sT/2}) \right)_s = -1$$

Search along the damping line until the angles add up to 180 degrees

$$s = -0.4922 + j0.9843$$

$$z = 0.6891 + j0.3695$$

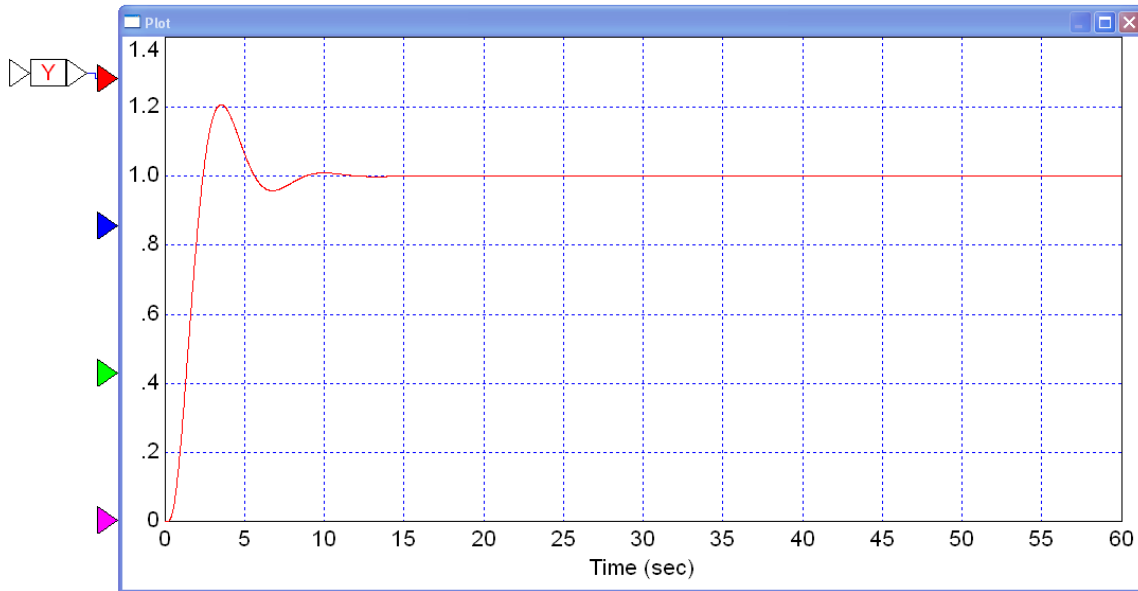
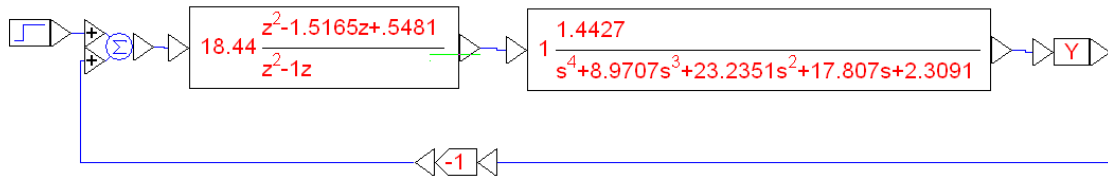
resulting in

$$\left(\left(\frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right) \cdot \left(\frac{k(z-0.922)(z-0.5945)}{z(z-1)} \right) \cdot (e^{-sT/2}) \right)_s = 0.0542 \angle 180^\circ$$

k is then

$$k = \frac{1}{0.0542} = 18.4478$$

$$K(z) = 18.4478 \left(\frac{(z-0.922)(z-0.5945)}{z(z-1)} \right)$$



Meeting Design Specs

4) Design a digital controller with $T = 0.5$ seconds that results in

- No error for a step input
- 20% overshoot for the step response, and
- A 2% settling time of 10 seconds

Simulate the step response of the closed-loop system (VisSim or Simulink preferred with $K(z)*G(s)$)

Translation

- Make the system type 1
- Place the closed-loop dominant pole at $s = -0.4 + j0.8$
- Place the closed-loop dominant pole at $z = 0.7541 + j0.3188$

Let

$$K(z) = k \left(\frac{(z-0.922)(z-0.5945)}{(z-1)(z-a)} \right)$$

Evaluate what we know

$$\left(\left(\frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right) \cdot \left(\frac{(z-0.922)(z-0.5945)}{(z-1)} \right) \cdot (e^{-sT/2}) \right)_s$$

$$= 0.0514 \angle -144.64^\circ$$

The pole at $z = a$ must contribute the remainder (-35.3567 degrees)

$$\angle(z-a) = 35.3567^\circ$$

$$a = 0.7571 - \frac{0.3188}{\tan(35.3567^\circ)} = 0.3078$$

and

$$K(z) = k \left(\frac{(z-0.922)(z-0.5945)}{(z-1)(z-0.3078)} \right)$$

Evaluating at the design point (s and z):

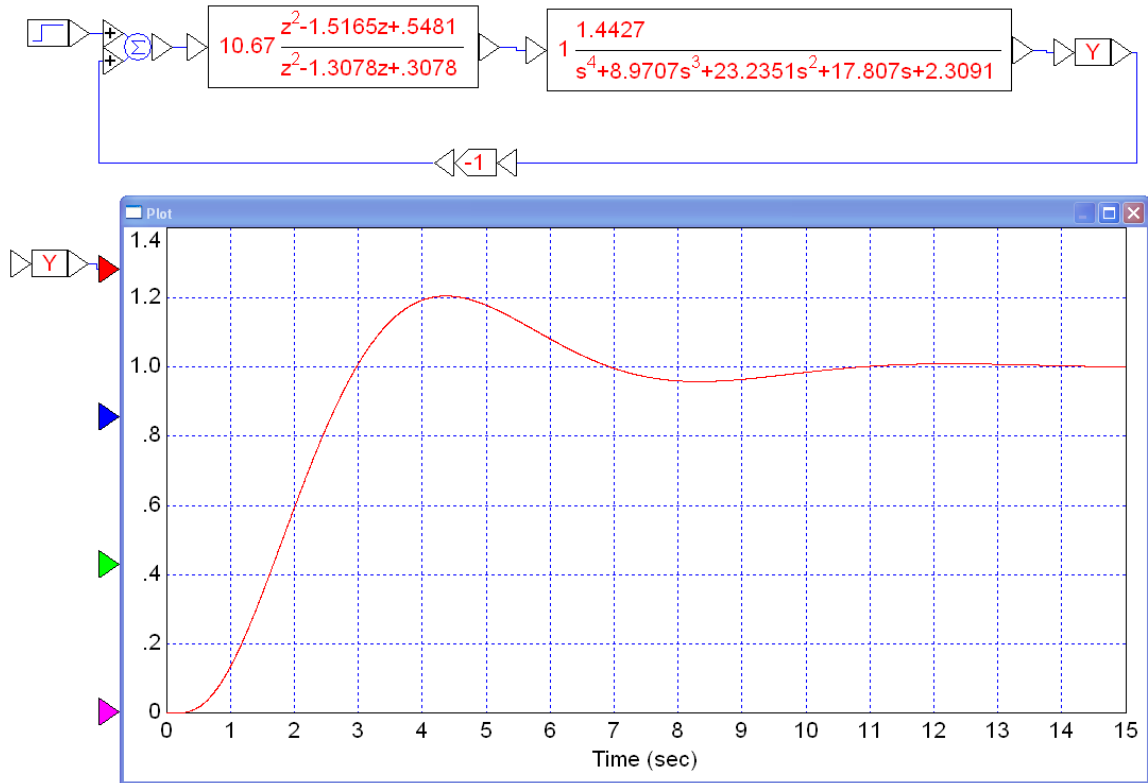
$$\left(\left(\frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right) \cdot \left(\frac{(z-0.922)(z-0.5945)}{(z-1)(z-0.3078)} \right) \cdot (e^{-sT/2}) \right)_s$$

$$= 0.0937 \angle 180^\circ$$

meaning

$$k = \frac{1}{0.0937} = 10.671$$

$$K(z) = 10.67 \left(\frac{(z-0.922)(z-0.5945)}{(z-1)(z-0.3078)} \right)$$



5) Design a digital controller with $T = 0.1$ second that results in

- No error for a step input
- 20% overshoot for the step response, and
- A 2% settling time of 4 seconds

Simulate the step response of the closed-loop system (VisSim or Simulink preferred with $K(z)*G(s)$)

Translation

- Make the system type 1
- Place the closed-loop dominant pole at $s = -0.4 + j0.8$
- Place the closed-loop dominant pole at $z = 0.9577 + j0.0768$

Let

$$K(s) = k \left(\frac{(s+0.1617)(s+1.04)}{s(s+b)} \right)$$

$$K(z) = k \left(\frac{(z-0.9940)(z-0.9012)}{(z-1)(z-a)} \right)$$

Evaluate what we know

$$\left(\left(\frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right) \cdot \left(\frac{(z-0.9940)(z-0.9012)}{(z-1)} \right) \cdot (e^{-sT/2}) \right)_s$$

$$= 0.137 \angle -138.92^\circ$$

The pole at $(z = a)$ must subtract 41.08 degrees to bring this to 180 degrees

$$\angle(z + a) = 41.08^\circ$$

$$a = 0.9577 - \frac{0.0768}{\tan(41.08^\circ)} = 0.8696$$

so now

$$K(z) = k \left(\frac{(z-0.9940)(z-0.9012)}{(z-1)(z-0.8696)} \right)$$

Evaluate what we now know

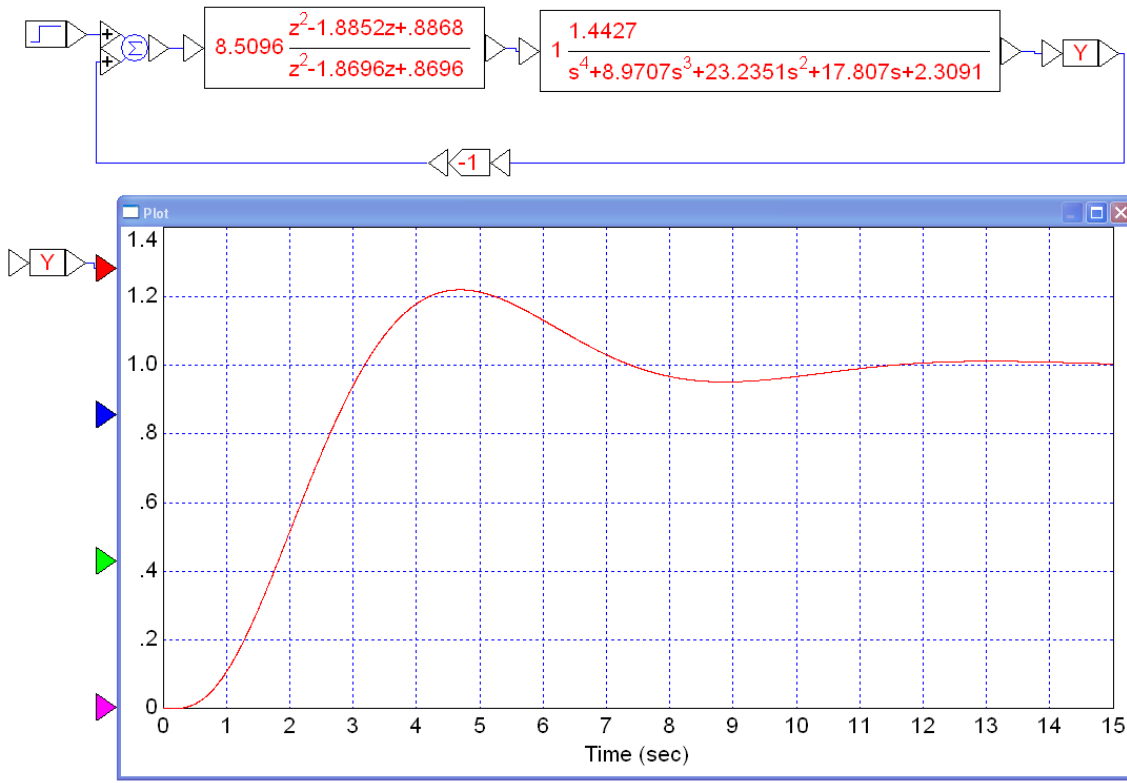
$$\left(\left(\frac{1.4427}{(s+0.1617)(s+1.04)(s+2.719)(s+5.05)} \right) \cdot \left(\frac{(z-0.9940)(z-0.9012)}{(z-1)(z-0.8696)} \right) \cdot (e^{-sT/2}) \right)_s$$

$$= 0.1175 \angle 180^\circ$$

k is then

$$k = \frac{1}{0.1175} = 8.5096$$

$$K(z) = 8.5096 \left(\frac{(z-0.9940)(z-0.9012)}{(z-1)(z-0.8696)} \right)$$



Note: If you change the sampling rate, it's a complete re-design.