
Root Locus for Systems with Delays

ECE 461/661 Controls Systems

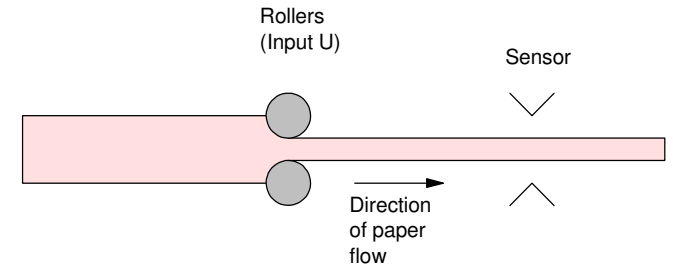
Jake Glower - Lecture #27

Please visit [Bison Academy](#) for corresponding
lecture notes, homework sets, and solutions

Source of Delays

Delays can result from many situations:

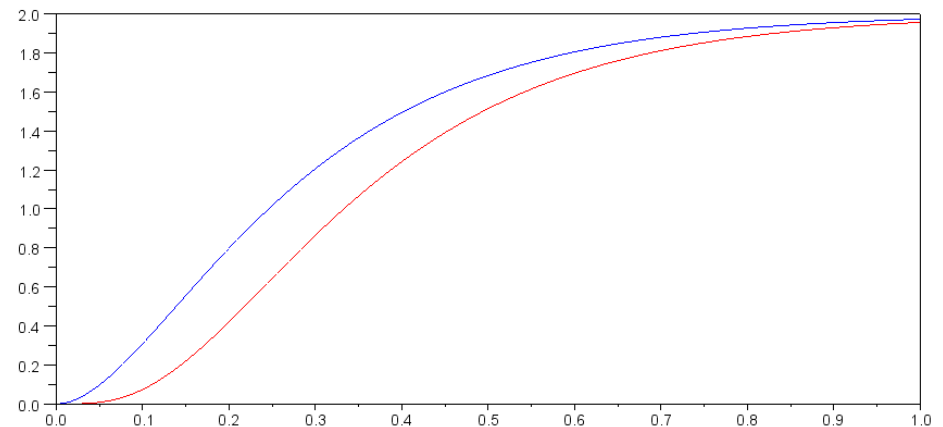
- Placing a sensor far from the input
Example: paper mill
- Ignored dynamics
fast poles do have some affect on a system
A delay is a way to model the poles you're ignoring



Example

$$G(s) = \left(\frac{60,000}{(s+5)(s+10)(s+20)(s+30)} \right)$$

$$G(s) \approx \left(\frac{100}{(s+5)(s+10)} \right)$$

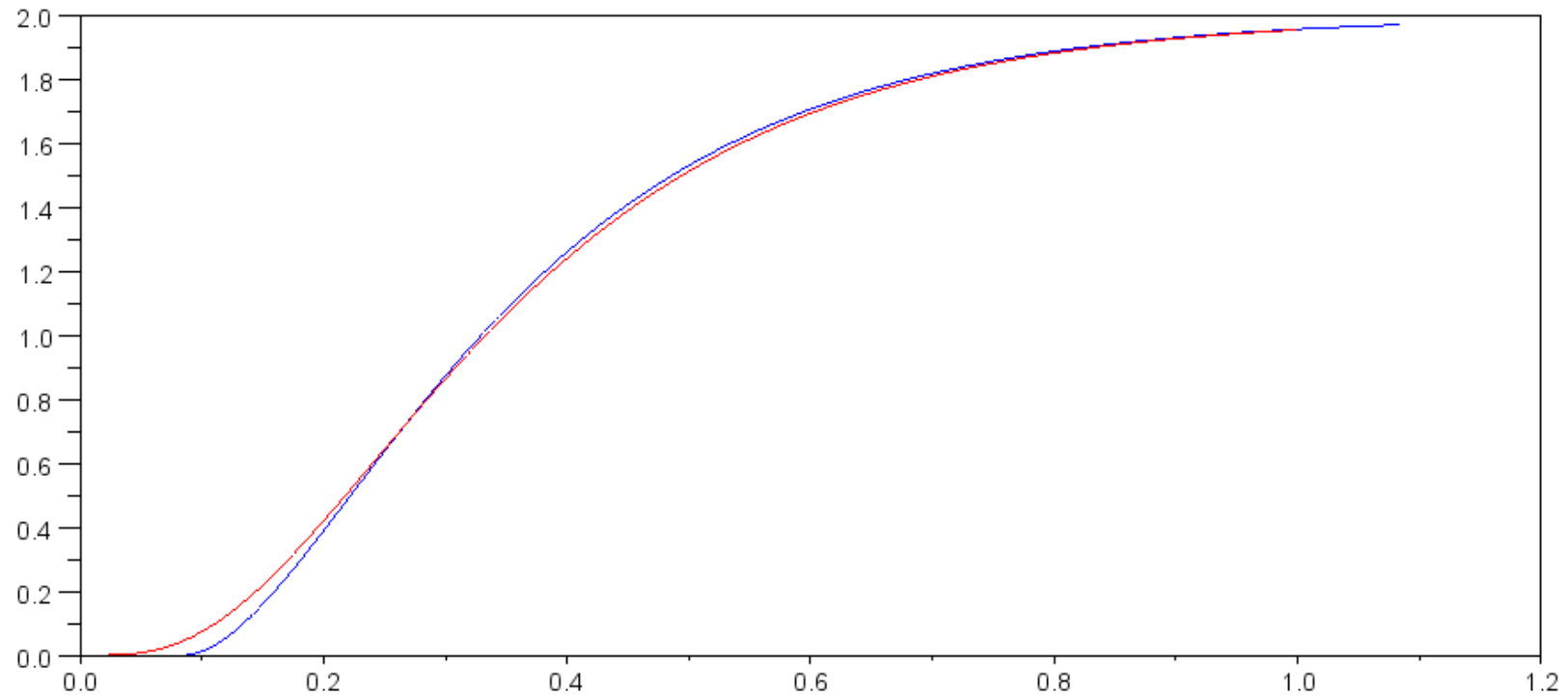


Approximating the delay

$$G(s) = \left(\frac{60,000}{(s+5)(s+10)(s+20)(s+30)} \right) \approx \left(\frac{100}{(s+5)(s+10)} \right) \cdot e^{-sT}$$

Find T to match the phase shift at $s = j1$

$$(e^{-sT})_{s=j1} = 1 \angle -4.7716^\circ \quad \Rightarrow \quad T = 0.0833 \text{ seconds}$$



Pade Approximation:

To use root locus techniques, we need the poles and zeros:

$$\text{Delay}(T) = e^{-sT} \approx k \frac{z(s)}{p(s)}$$

Rewrite as

$$e^{-sT} = \left(\frac{e^{-\frac{sT}{2}}}{e^{\frac{sT}{2}}} \right)$$

Expand as a Taylor's series

$$e^{-sT} = \left(\frac{1 - \left(\frac{T}{2}\right)s + \left(\frac{T^2}{8}\right)s^2 - \left(\frac{T^3}{48}\right)s^3 + \left(\frac{T^4}{384}\right)s^4 + \dots}{1 + \left(\frac{T}{2}\right)s + \left(\frac{T^2}{8}\right)s^2 + \left(\frac{T^3}{48}\right)s^3 + \left(\frac{T^4}{384}\right)s^4 + \dots} \right)$$

The more terms you add, the better the approximation.

Example: Find 'k' for 20% overshoot

$$Y = \left(\left(\frac{100}{s(s+5)(s+10)} \right) \cdot e^{-0.5s} \right) U$$

Solution #1: Use a Pade approximation with 2 terms: (Matlab function)

```
[num, den] = pade(0.5, 2);  
Delay = tf(num, den);
```

$$e^{-0.5s} \approx \left(\frac{s^2 - 12s + 48}{s^2 + 12s + 48} \right) = \left(\frac{(s-6+j3.464)(s-6-j3.464)}{(s+6+j3.464)(s+6-j3.464)} \right)$$



In Matlab:

```
G = zpk([], [0, -5, -10], 100);  
[num, den] = pade(0.5, 2);  
Delay = tf(num, den);  
k = logspace(-2, 2, 1000)';  
R = rlocus(G*Delay, k);
```

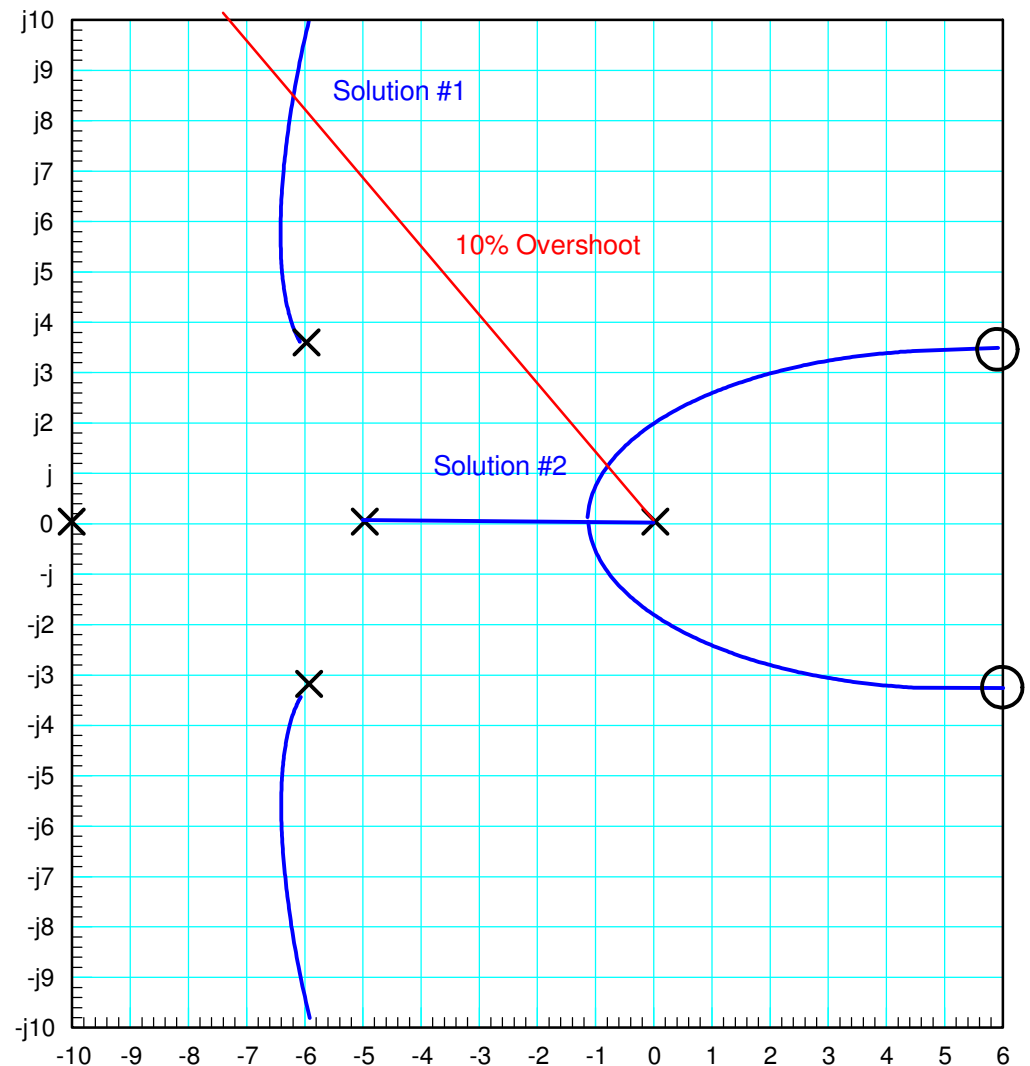
Two solutions:

$$s = -0.6646 + j1.3293$$

- $k = 0.4484$
- dominant pole

$$s = -4.5751 + j9.1502$$

- $k = 4.3299$
- stray solution



Checking in Matlab

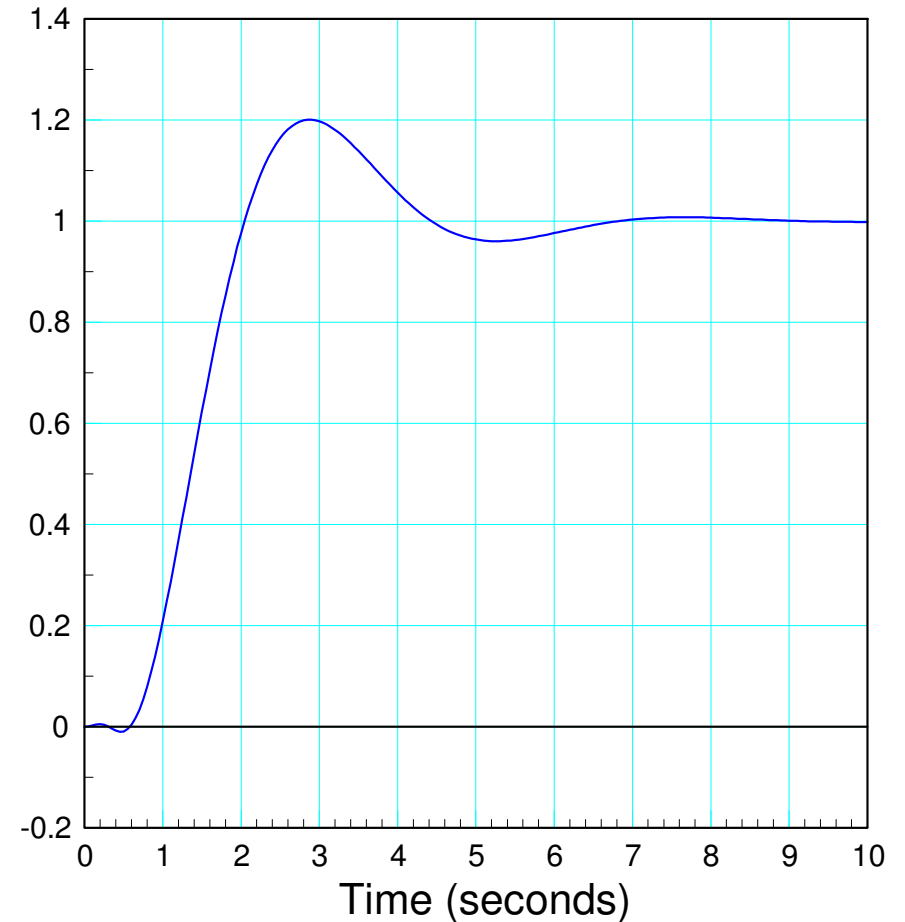
- Wobble for $0 < t < 1/2$ second due to Pade approximation:

```
G = zpk([], [0, -5, -10], 100);  
[num, den] = pade(0.5, 2);  
Delay = tf(num, den)
```

$$\frac{s^2 - 12s + 48}{s^2 + 12s + 48}$$

```
k = 0.4484;
```

```
Gcl = G*Delay*k / (1 + G*Delay*k);  
Gcl = minreal(Gcl);  
t = [0:0.01:10]';  
y = step(Gcl, t);  
plot(t, y);
```

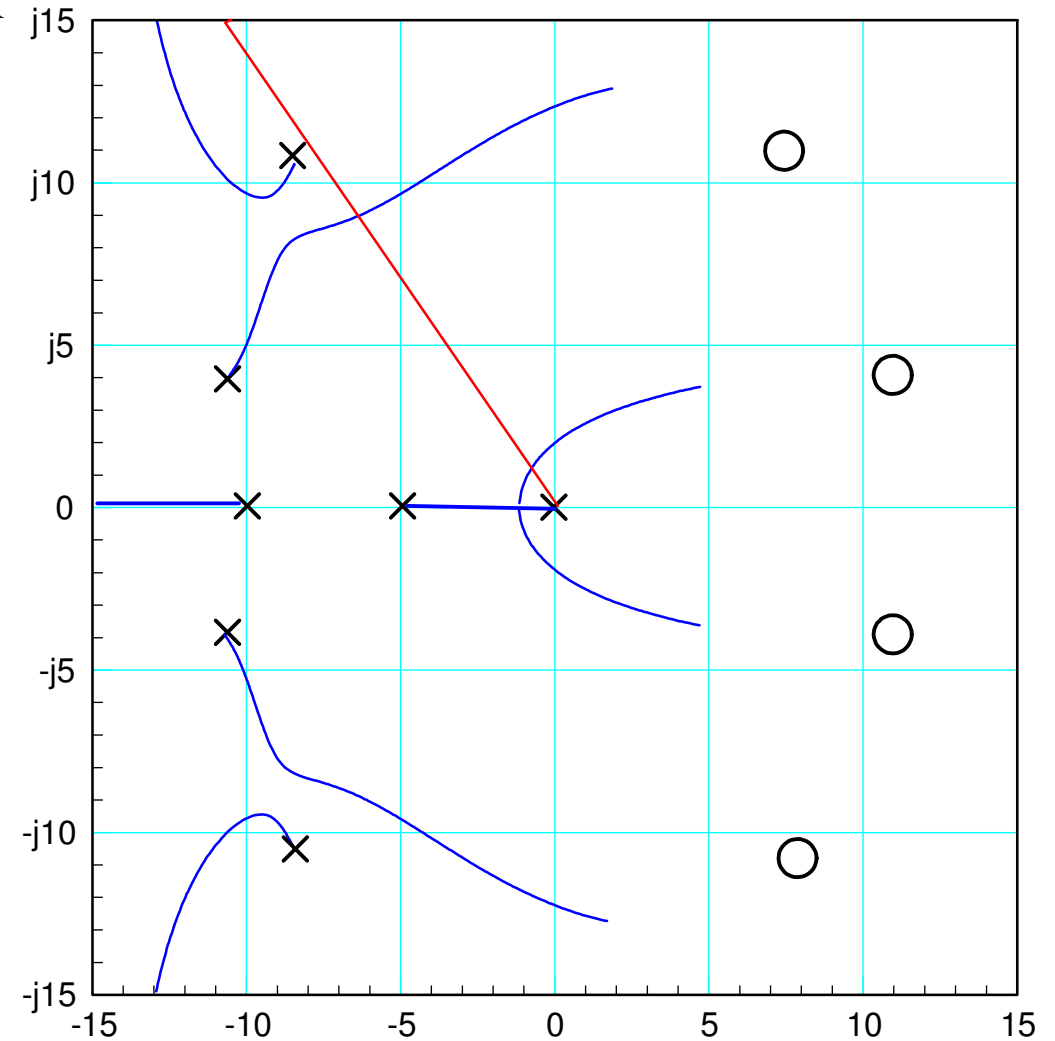


Problem: 4th-order Pade Approximation

$$e^{-0.5s} \approx \left(\frac{(s-8.42 \pm j10.63)(s-11.58 \pm j3.47)}{(s+8.42 \pm j10.63)(s+11.58 \pm j3.47)} \right)$$

Root Locus:

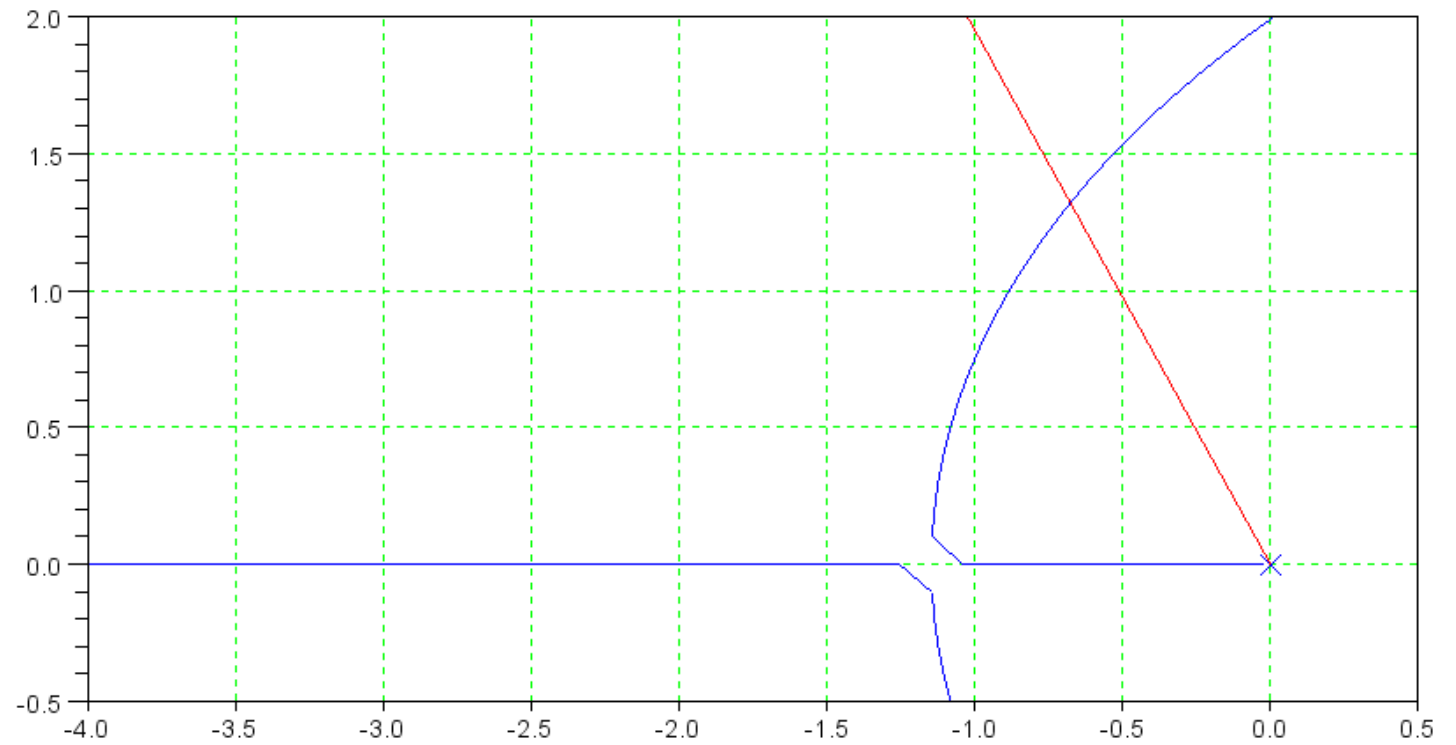
```
[num,den] = pade(0.5, 4);  
Delay = tf(num,den)  
k = logspace(-2,2,1000)';  
G = zpk([], [0,-5,-10], 100)  
rlocus(G*Delay,k);
```



Zooming in on the dominant pole:

$$s = -0.6666 + j1.3333$$

$$k = 0.4566$$



Option #2: Numerical Solution

Any point on the root locus satisfies

$$\angle(GK) = 180^0$$

Search along the damping line until the angles add up 180 degrees

For 20% overshoot

$$\text{angle}\left(\frac{100 \cdot e^{-0.5s}}{s(s+5)(s+10)}\right)_{s=\alpha(-1+j2)} = 180^0$$

```
s = -1 + j*2;  
G = zpk([], [0, -5, -10], 100);  
evalfr(G, s) * exp(-0.5*s)
```

```
ans = -1.5006 + 0.9728i
```

```
s = s * 0.9;  
evalfr(G, s) * exp(-0.5*s)
```

```
ans = -1.7267 + 0.7338i
```

This method doesn't have a name and might not be taught anywhere else other than NDSU.

- It isn't really root locus design, since you're not drawing the root locus
 - It sort of is root locus design, since you're finding the point on the root locus which intersects the desired damping line. That's the only point you care about anyway, so you don't need (and won't use) the rest of the root loci.
 - It's a lot easier and more accurate since you use e^{-sT} to model a delay, not an approximation. (Typing in four poles and four zeros for the 4th-order model was a bit of a pain. Typing in e^{-sT} was easy.)
-

In Matlab, iterate until the angle of $G * \text{delay}$ is zero. Taking your initial guess as

$$s = -0.5 + j$$

then iterating by scaling s each step results in

```
s = 0.5 * (-1 + j*2);  
evalfr(G,s) * exp(-s*T)
```

- 2.5039068 - 0.7297865i

```
s = s * 1.1;           10% larger  
evalfr(G,s) * exp(-s*T)
```

- 2.4062416 - 0.4851018i *good: complex portion is getting smaller.*

```
s = s * 1.1;           add another 10%  
evalfr(G,s) * exp(-s*T)
```

- 2.303687 - 0.2437231i *better: complex portion is getting smaller*

```
s = s * 1.1;           add another 10%  
evalfr(G,s) * exp(-s*T)
```

- 2.1923504 - 0.0045661i *close: complex portion is almost zero*

(time passes)

$$s = 0.666717 * (-1 + j*2)$$

$$- 0.666717 + 1.333434i$$

$$\text{evalfr}(G, s) * \exp(-s*T)$$

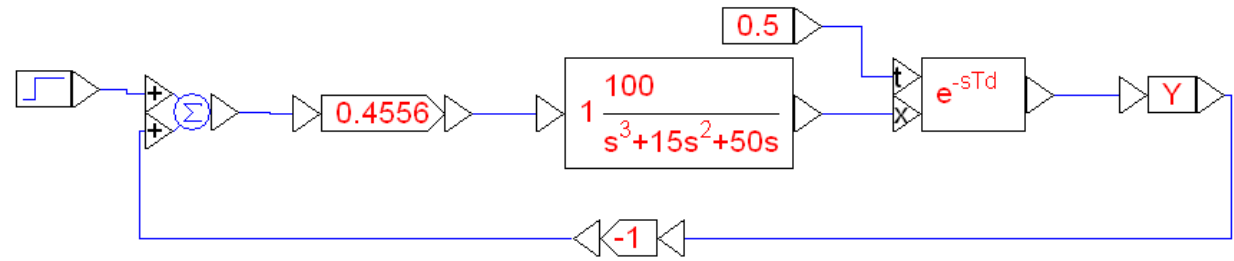
$$- 2.1901017 + 0.0000002i \quad \text{close enough}$$

If you use this method, the result is the exact solution

Method	s	k
2nd-Order Pade	-0.6646 + j1.3293	0.4484
4th-Order Pade	-0.6666 + j1.3333	0.4566
exp(-sT) (exact)	-0.6667 + j1.3334	0.4566

Positives

- Works
- Doesn't need Pade approximations
- Exact solution



Negatives

- Not a built in function in Matlab

