
Converting $G(s)$ to $G(z)$

ECE 461/661 Controls Systems

Jake Glower - Lecture #30

Please visit [Bison Academy](#) for corresponding
lecture notes, homework sets, and solutions

Digital Control Systems

Often times, the compensator, $K(s)$, will be implemented with a microcontroller.

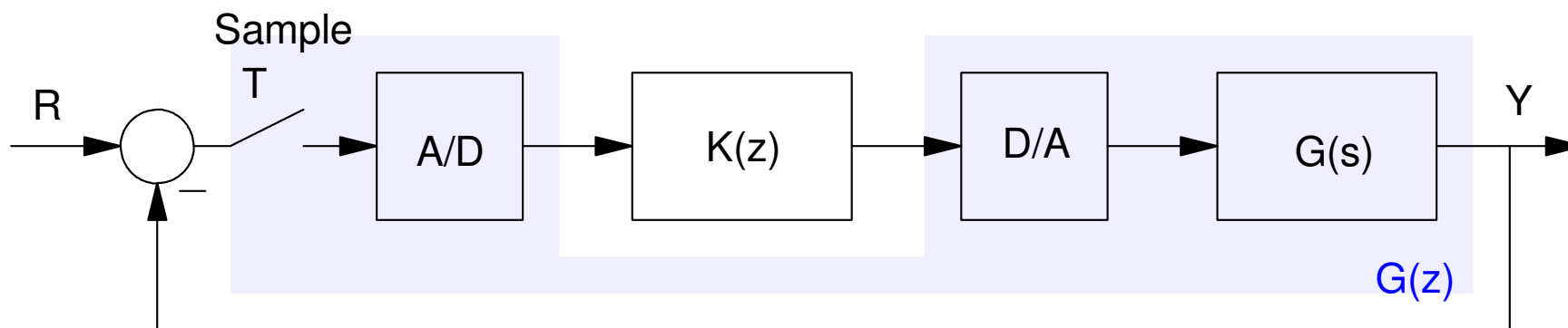
- The result is a hybrid system: both analog and discrete

Makes analysis difficult

Relative to the microcontroller, the world looks discrete:

- You output a control signal every T seconds,
- Through the system dynamics, this results in an error, sampled every T seconds as well.
- Convert everything to the z -domain

Now you can analyze the closed-loop system



Converting $G(s)$ to $G(z)$

Method 1: Substitution

- sY means *the derivative of Y*

Euler - Backward Difference

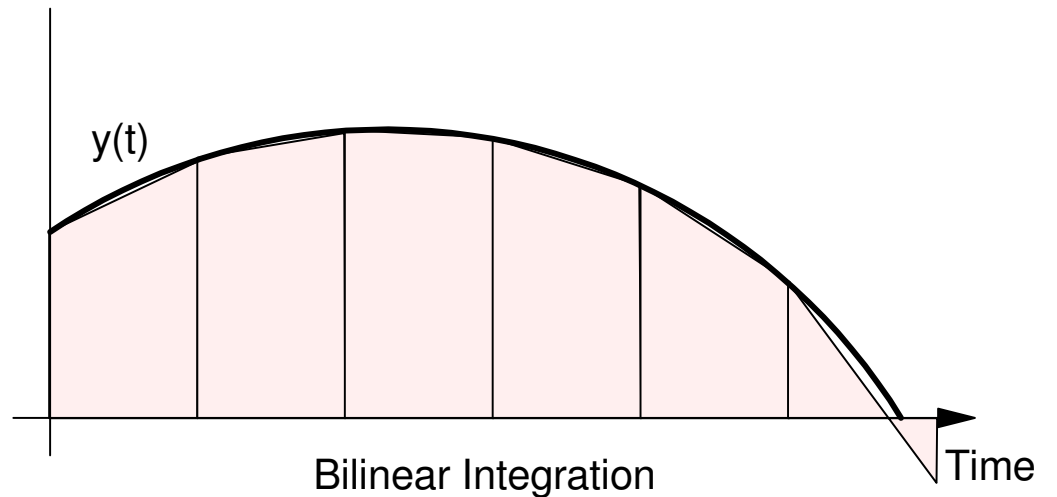
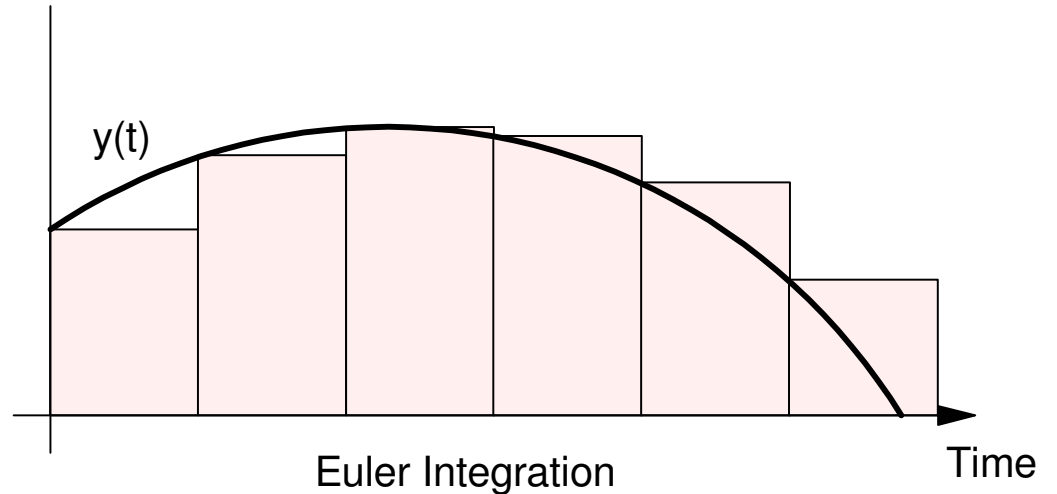
$$s \approx \left(\frac{z-1}{T} \right)$$

Euler - Forward Difference:

$$s \approx \left(\frac{z-1}{Tz} \right)$$

Bilinear

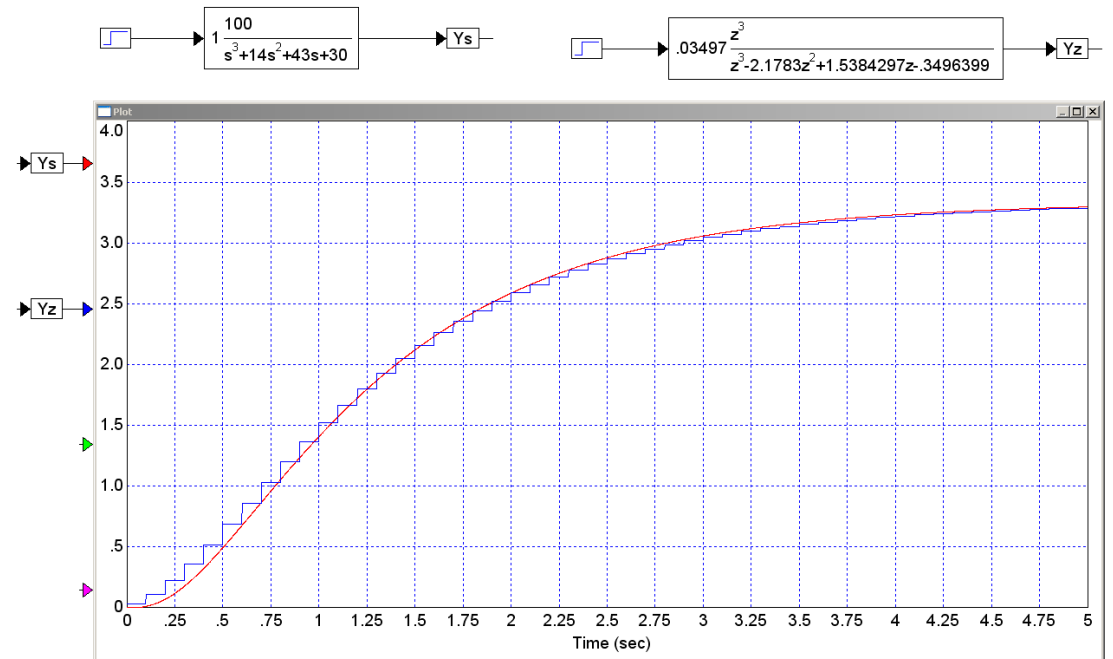
$$s \approx \frac{2}{T} \left(\frac{z-1}{z+1} \right)$$



Example: Find the z-transform of $G(s)$

- Assume a sampling rate of 100ms ($T = 0.1$).

$$G(s) = \left(\frac{100}{(s+1)(s+3)(s+10)} \right)$$



a) Using Euler Forward Difference:

$$G(s) = \left(\frac{100}{(s+1)(s+3)(s+10)} \right)$$

$$G(z) \approx \left(\frac{100}{\left(\left(\frac{z-1}{Tz} \right) + 1 \right) \left(\left(\frac{z-1}{Tz} \right) + 3 \right) \left(\left(\frac{z-1}{Tz} \right) + 10 \right)} \right)$$

$$G(z) = \left(\frac{0.03497z^3}{(z-0.9091)(z-0.7692)(z-0.5)} \right)$$

In Matlab: Input the system in the s-plane

```
Gs = zpk([], [-1, -3, -10], 100)
```

$$\frac{100}{(s+1)(s+3)(s+10)}$$

If you add one more term, Matlab interprets this as a discrete-time system with the last term being the sampling rate:

```
Gz = zpk([0, 0, 0], [0.9091, 0.7692, 0.5], 0.03497, 0.1)
```

$$\frac{0.03497 z^3}{(z-0.9091)(z-0.7692)(z-0.5)}$$

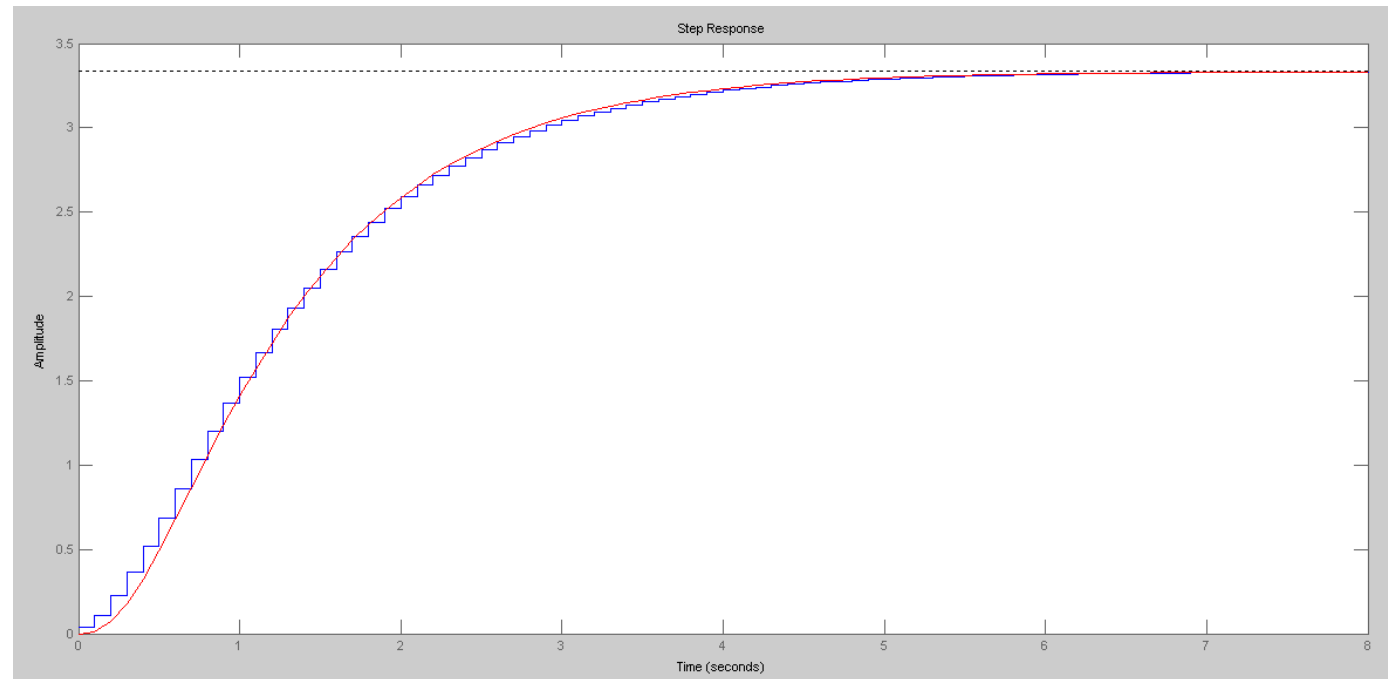
```
Sampling time (seconds): 0.1
```

To plot the step response of the two systems together

- Plot the step response of $G(z)$
- Type *hold on* to keep this plot
- Plot the step response of $G(s)$ on top of the $G(z)$ graph

```
step(Gz)
```

```
t = [0:0.01:8]';  
ys = step(Gs,t);  
hold on  
plot(t,ys,'r')
```

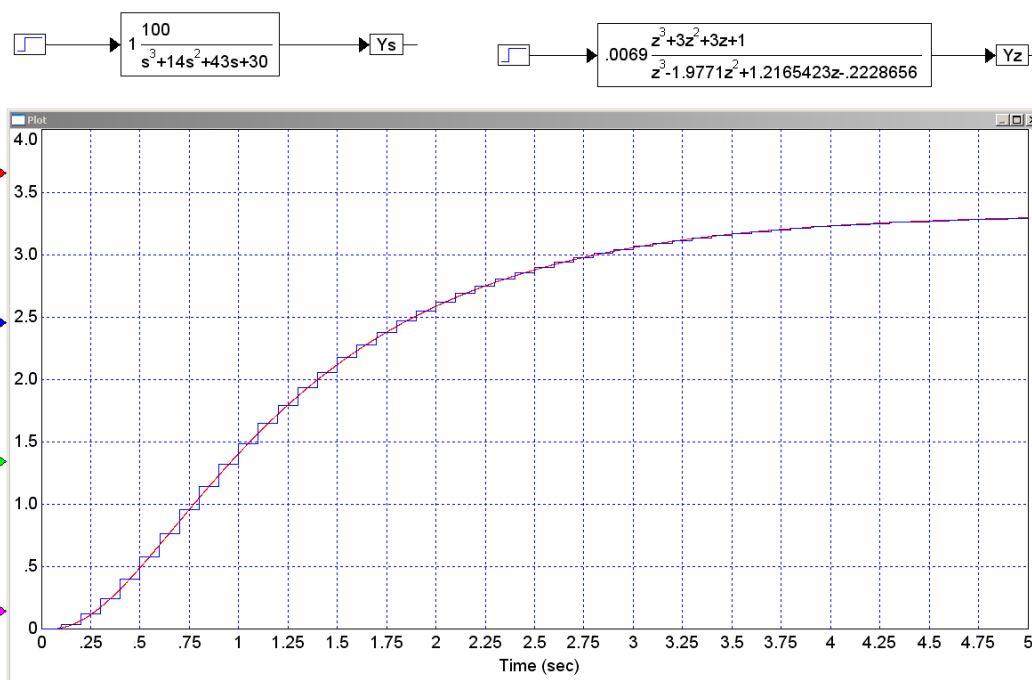


Bilinear:

$$G(s) = \left(\frac{100}{(s+1)(s+3)(s+10)} \right)$$

$$G(z) \approx \left(\frac{100}{\left(\frac{2}{T}\left(\frac{z-1}{z+1}\right)+1\right)\left(\frac{2}{T}\left(\frac{z-1}{z+1}\right)+3\right)\left(\frac{2}{T}\left(\frac{z-1}{z+1}\right)+10\right)} \right)$$

$$G(z) \approx \left(\frac{0.00690(z+1)^3}{(z-0.9047)(z-0.7391)(z-0.3333)} \right)$$



Method #2: Transform Poles and Zeros (my preference)

LaPlace transforms assume

$$y = e^{st}$$

z-transforms assume

$$y = z^k$$

Assume

$$t = kT$$

$$y = e^{s(kT)} = e^{(sT)k} = (e^{sT})^k = z^k$$

The conversion from the s-plane to the z-plane is

$$z = e^{sT}$$

Procedure:

i) Convert every pole and zero as

$$z = e^{sT}$$

ii) Add a gain to match the DC gain

iii) (optional) Add n zeros at $z = 0$ to match the phase at a frequency close to zero

- or -

Add n zeros at $z = 0$ to match the delay in the system.

Example: $G(z)$. Assume $T = 0.1$

$$G(s) = \left(\frac{100}{(s+1)(s+3)(s+10)} \right)$$

In Matlab

```
s = [-1, -3, -10]
```

```
    -1.00    -3.00   -10.00
```

```
T = 0.1;
```

```
z = exp(s*T)
```

```
    0.9048374    0.7408182    0.3678794
```

Meaning

$$G(z) = \left(\frac{k}{(z-0.9048)(z-0.7408)(z-0.3687)} \right)$$

Match the DC gain

$$\left(\frac{100}{(s+1)(s+3)(s+10)} \right)_{s=0} = 3.3333$$

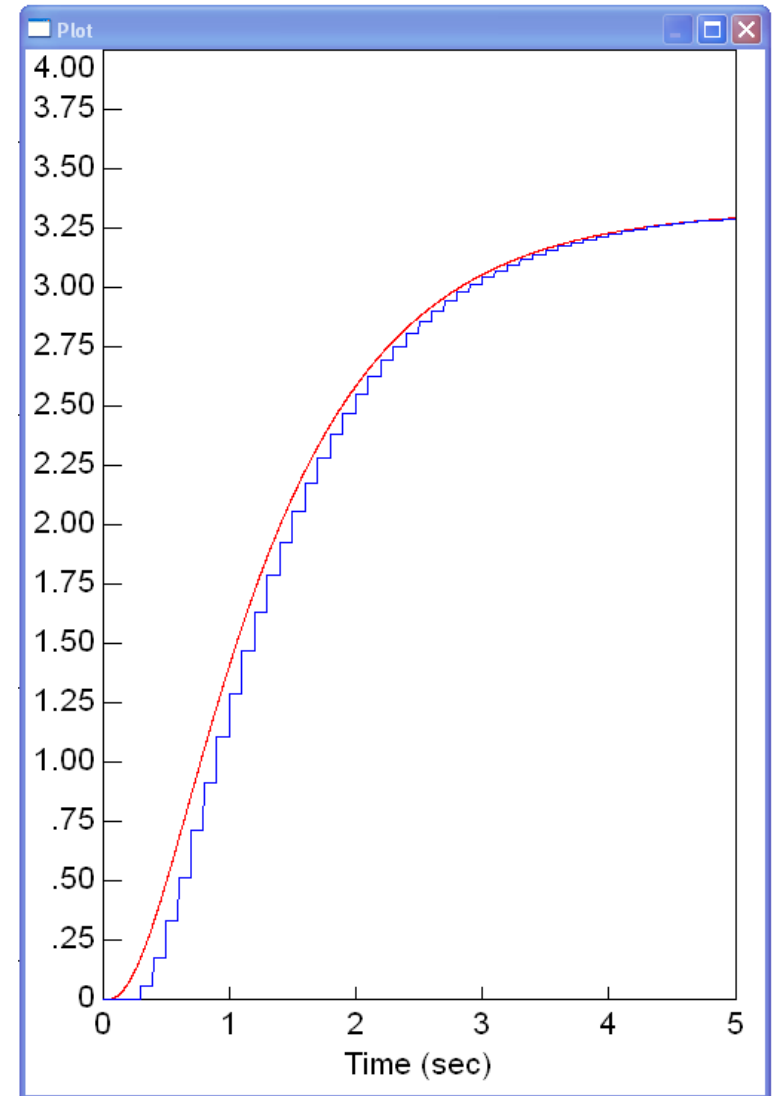
$$\left(\frac{k}{(z-0.9048)(z-0.7408)(z-0.3687)} \right)_{z=1} = 3.33333$$

$$k = \text{prod}(1-z) * 3.3333$$

$$0.0591691$$

meaning

$$G(z) = \left(\frac{0.059169}{(z-0.9048)(z-0.7408)(z-0.3687)} \right)$$



To find how many zeros belong at $z=0$,

a) There is too much delay with this system. Adjust by adding zeros at $z = 0$

b) Match the phase at some frequency, such as $s = j1$

$$\left(\frac{100}{(s+1)(s+3)(s+10)} \right)_{s=j1} = 2.2249 \angle -69.14^\circ$$

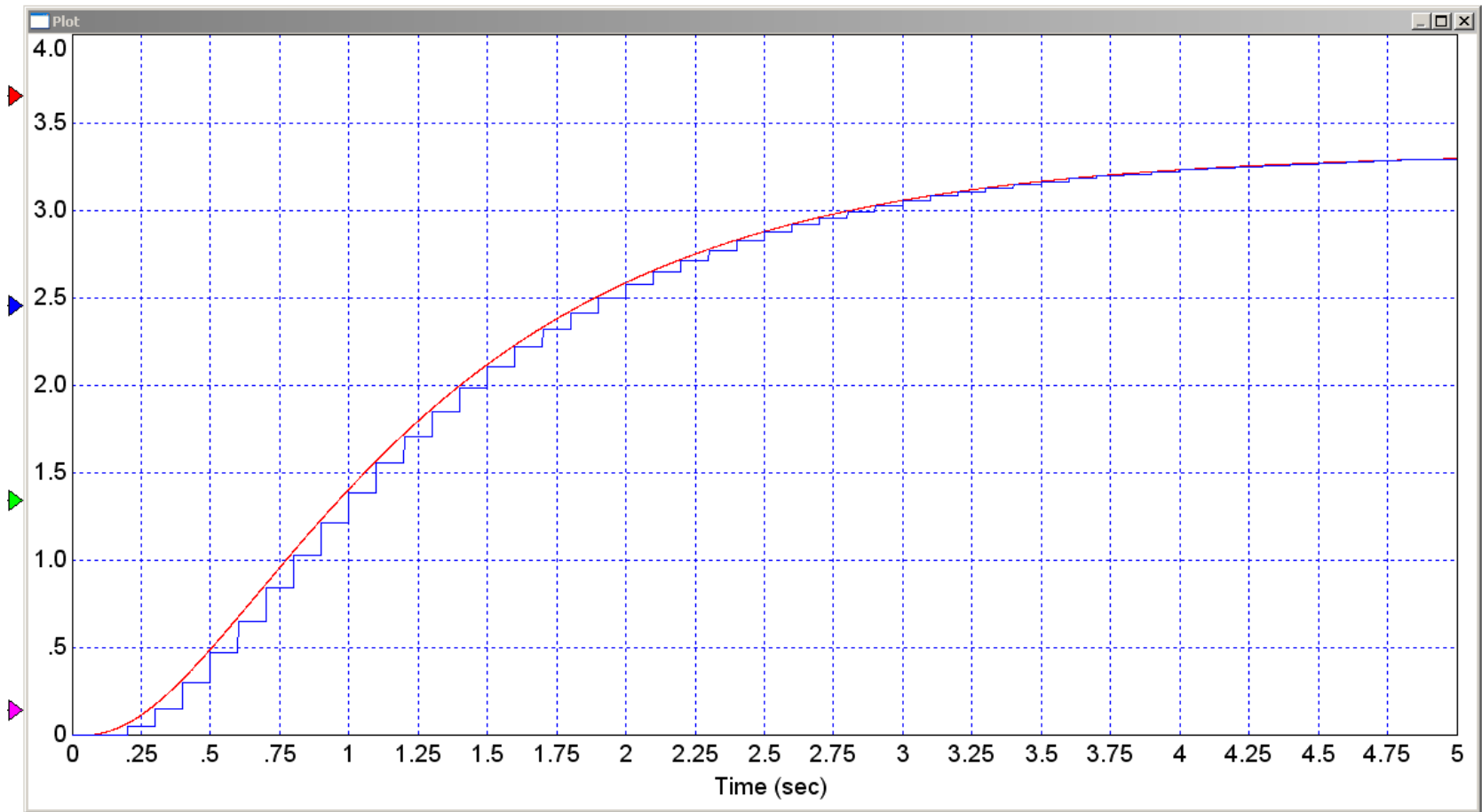
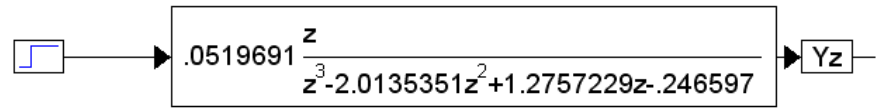
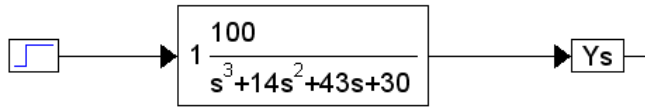
$$\left(\frac{0.051969}{(z-0.9048)(z-0.7408)(z-0.3687)} \right)_{s=j} = 2.2276 \angle -78.04^\circ$$

$$z = e^{sT} = e^{(j1)(0.1)} = 1 \angle 5.73^\circ$$

Add 1.55 zeros at $z=0$ to make the phase match

- Round up to 2 or round down to 1

$$G(z) \approx \left(\frac{0.051969 \cdot z}{(z-0.9048)(z-0.7408)(z-0.3687)} \right)$$



In Matlab: Input the system G(s)

```
Gs = zpk([], [-1, -3, -10], 100)
```

$$\frac{100}{(s+1)(s+3)(s+10)}$$

Input G(z). For now, assume the numerator is 1

```
T = 0.1;  
Gz = zpk([], [exp(-1*T), exp(-3*T), exp(-10*T)], 1, T)
```

$$\frac{1}{(z-0.9048)(z-0.7408)(z-0.3679)}$$

```
Sampling time (seconds): 0.1
```

Add a gain, k, so that the DC gain matches up

```
DCs = evalfr(Gs, 0)
```

```
3.3333
```

```
DCz = evalfr(Gz, 1)
```

```
64.1401
```

```
k = DCs / DCz
```

```
0.0520
```

So, $G(z)$ is.

```
Gz = zpk([], [exp(-1*T), exp(-3*T), exp(-10*T)], k, T)
```

```
0.05197
```

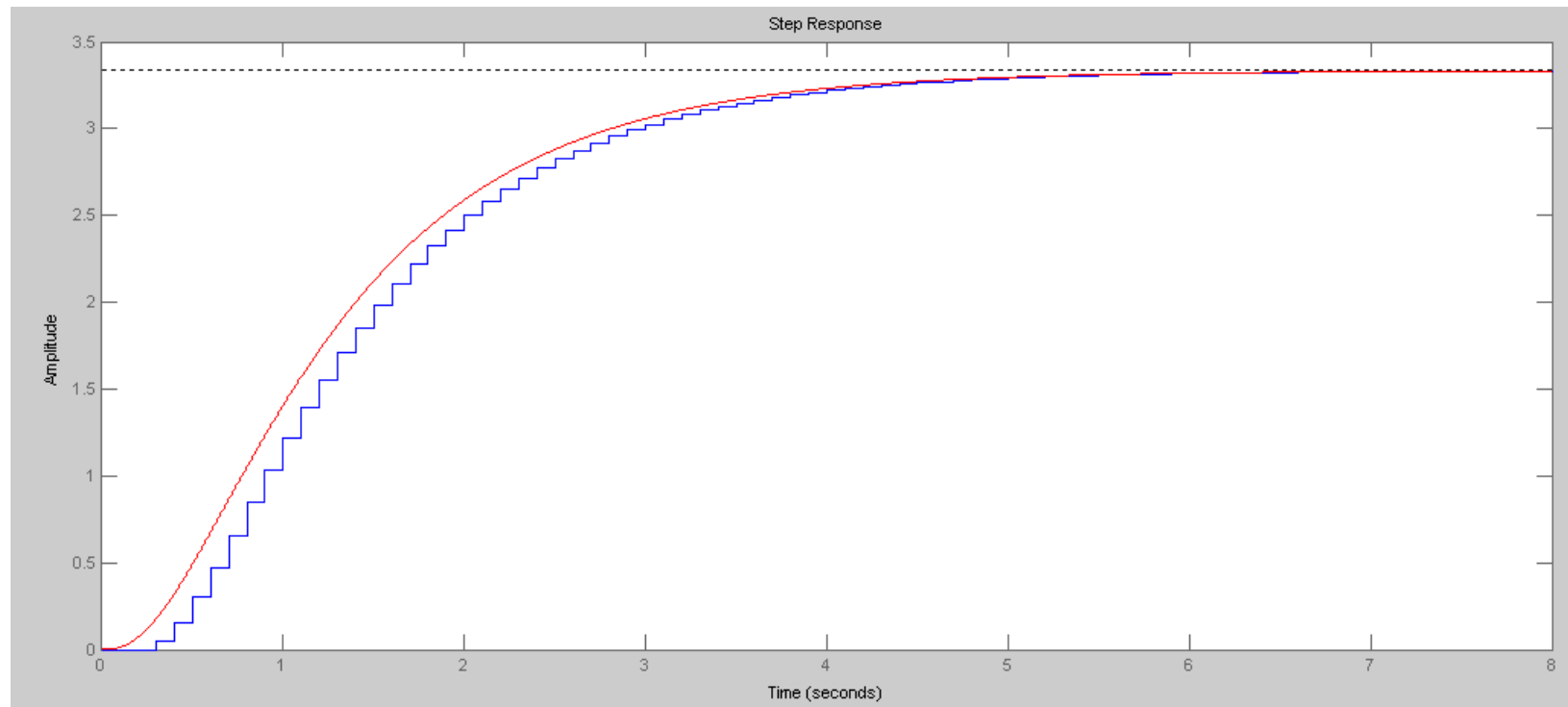
```
-----  
(z-0.9048) (z-0.7408) (z-0.3679)
```

```
Sampling time (seconds): 0.1
```

Checking the answer: Plot the step response of $G(z)$ and $G(s)$

- Add zeros at $z=0$ to remove the time delay

```
step(Gz)
hold on
t = [0:0.001:8]';
ys = step(Gs,t);
plot(t,ys,'r');
```



Example 2: Complex Poles

This also works with complex poles and zeros

$$G(s) = \left(\frac{100(s+j5)(s-j5)}{(s+1+j4)(s+1-j4)(s+20)} \right)$$

```
sn = [j*5, -j*5]'
```

```
T = 0.1
```

```
zn = exp(sns*T)
```

```
0.8775826 - 0.4794255i
```

```
0.8775826 + 0.4794255i
```

```
poly(nz)
```

```
1. - 1.7551651 1.
```

```
sd = [-1+j*4, -1-j*4, -20]'
```

```
- 1. - 4.i
```

```
- 1. + 4.i
```

```
- 20.
```

zd = exp(sd*T)

0.8334105 - 0.3523603i
0.8334105 + 0.3523603i
0.1353353

poly(zd)

1. - 1.8021562 1.0443104 - 0.1108032

meaning

$$G(z) = k \left(\frac{z^2 - 1.755z + 1}{z^3 - 1.802z^2 + 1.044z - 0.110} \right)$$



To find k, match the DC gain:

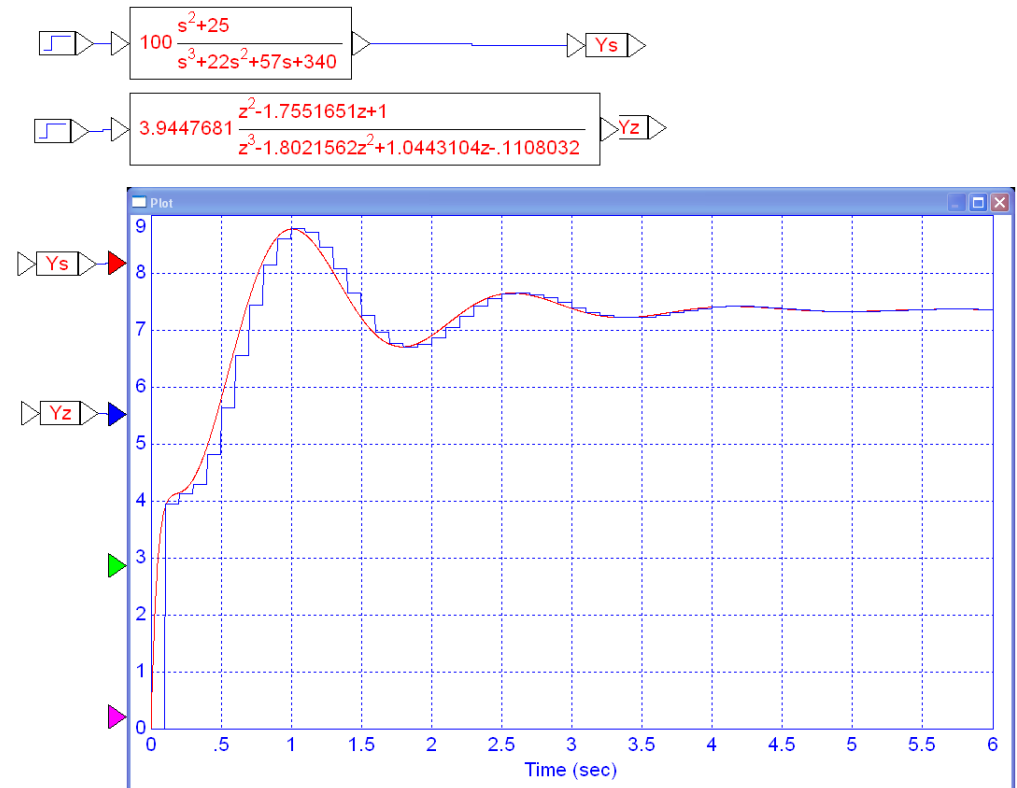
$$DC = 100 \cdot 25 / 340$$

$$7.3529412$$

$$k = DC \cdot \text{prod}(1-zd) / \text{prod}(1-zn)$$

$$3.9447681$$

$$G(z) = 3.944 \left(\frac{z^2 - 1.755z + 1}{z^3 - 1.802z^2 + 1.044z - 0.110} \right)$$



In Matlab: Input the system G(s)

```
z1 = j*5;  
z2 = -j*5;  
p1 = -1+j*4;  
p2 = -1-j*4;  
p3 = -2;  
Gs = zpks([z1, z2], [p1, p2, p3], 100)
```

$$\frac{100 (s^2 + 25)}{(s+2) (s^2 + 2s + 17)}$$

Now input G(z). Convert the poles and zeros to the z-plane as e^{sT} .

```
T = 0.1;  
Gz = zpks([exp(z1*T), exp(z2*T)], [exp(p1*T), exp(p2*T), exp(p3*T)], 1, T)
```

$$\frac{(z^2 - 1.755z + 1)}{(z-0.8187) (z^2 - 1.667z + 0.8187)}$$

Sampling time (seconds): 0.1

Add a gain, k, to make the DC gains match up:

```
DCs = evalfr(Gs, 0)
```

```
73.5294
```

```
DCz = evalfr(Gz, 1)
```

```
8.8913
```

```
k = DCs / DCz
```

```
8.2699
```

So, the discrete-time model for G(s) is....

```
Gz = zpk([exp(z1*T), exp(z2*T)], [exp(p1*T), exp(p2*T), exp(p3*T)], k, T)
```

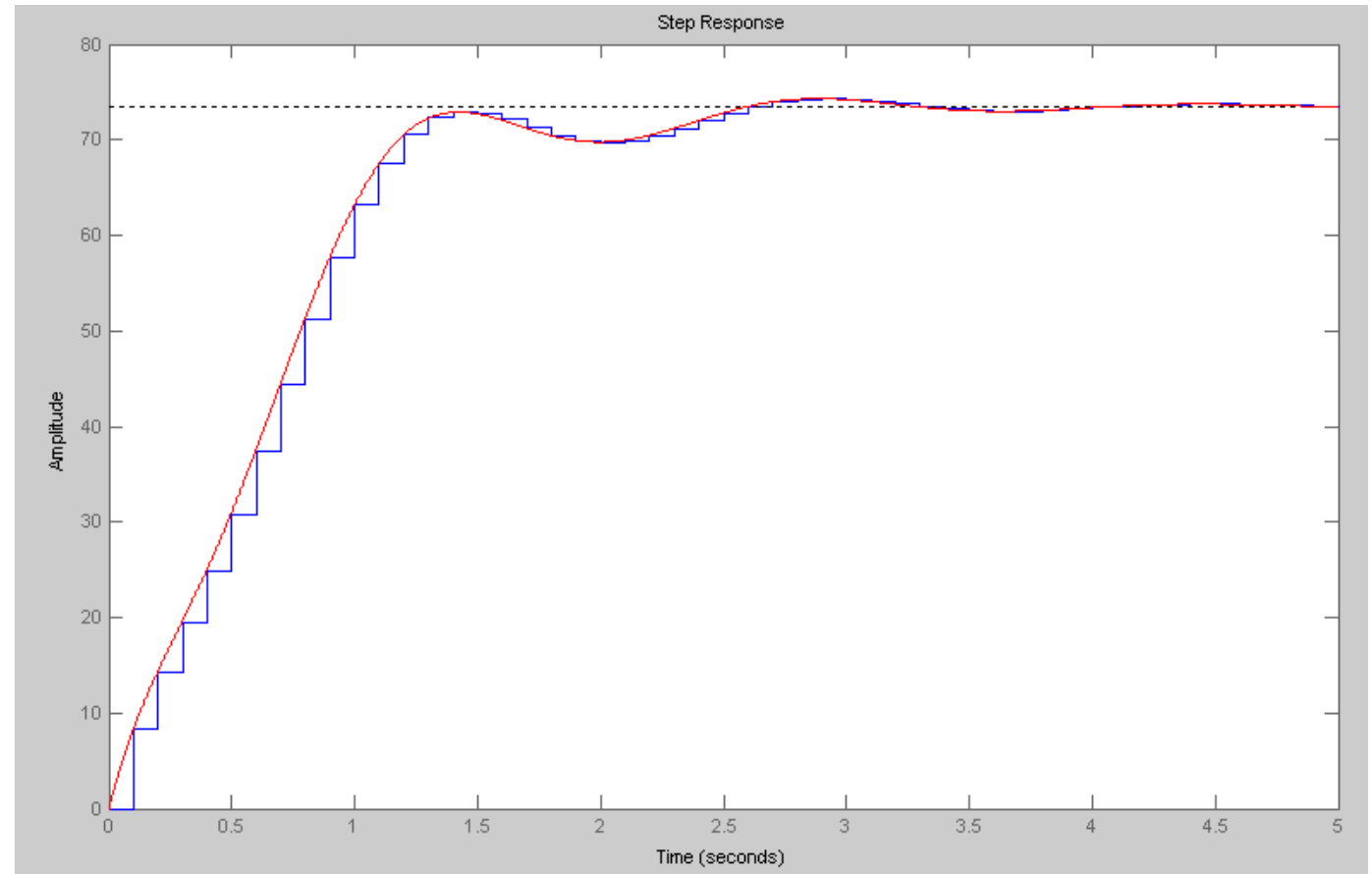
```
8.2699 (z^2 - 1.755z + 1)
```

```
-----  
(z-0.8187) (z^2 - 1.667z + 0.8187)
```

```
Sampling time (seconds): 0.1
```

Check the result by plotting the step response of $G(s)$ and $G(z)$ on the same graph:

```
step(Gz)
hold on
t = [0:0.001:5]';
ys = step(Gs,t);
plot(t,ys,'r');
```



Handout: Determine the discrete-time equivalent of $G(s)$. Assume $T = 0.1$ seconds

$$G(s) = \left(\frac{20}{(s+2)(s+5)} \right)$$



Handout: Determine the continuous-time equivalent of $G(z)$. Assume $T = 0.1$ second

$$G(z) = \left(\frac{0.2z}{(z-0.8)(z-0.6)} \right)$$



Note: Changing the Sampling Rate:

$$z = e^{sT}$$

If you change the sampling rate, $G(z)$ changes

and all of your analysis on $G(z)$ becomes worthless

$$G(s) = \left(\frac{100}{(s+1)(s+3)(s+10)} \right)$$

At $T = 0.1$

$$G(z) \approx \left(\frac{0.051969 \cdot z}{(z-0.9048)(z-0.7408)(z-0.3687)} \right)$$

At $T = 0.01$

$$G(z) \approx \left(\frac{0.00009393 \cdot z}{(z-0.9900)(z-0.9704)(z-0.9048)} \right)$$

Note 2: Frequency Response of $G(z)$

If $G(s)$ and $G(z)$ are the same system

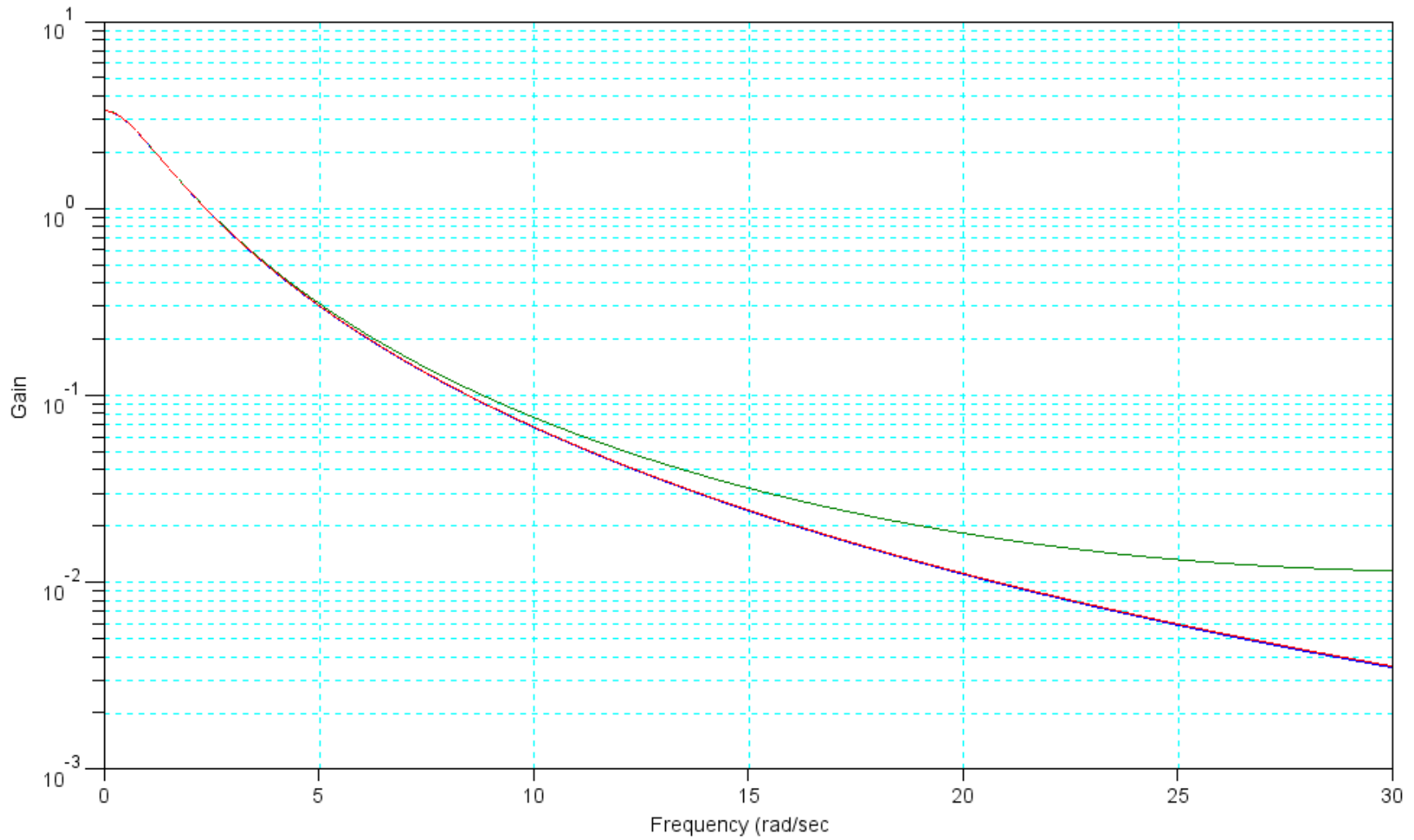
- They have the step response
- They have the same frequency response

```
w = [0:0.01:30]';
s = j*w;
Gs = 100 ./ ( (s+1) .* (s+3) .* (s+10) );

T = 0.1;
z = exp(s*T);
Gz = 0.051969*z ./ ( (z-0.9048) .* (z-0.7408) .* (z-0.3687) );

T = 0.01;
z = exp(s*T);
Gz2 = 0.000093938*z ./ ( (z-0.9900) .* (z-0.9704) .* (z-0.9048) );

plot(w, abs(Gs), w, abs(Gz), w, abs(Gz2));
xlabel('Frequency (rad/sec)');
ylabel('Gain');
xgrid(4)
```



Gain of $G(s)$ (blue), $G(z)$ with $T = 0.1$ (green), and $G(z)$ with $T = 0.01$ (red)

Summary

There are several ways to convert from $G(s)$ to $G(z)$

- Substitution:

Replace '1/s' with a numerical approximation for integration

- Mapping:

Map the poles and zeros to the z-plane as $z = e^{sT}$

Then match the DC gain

Either method works. As a check

- Both $G(s)$ and $G(z)$ should have similar step responses
- Both $G(s)$ and $G(z)$ should have similar frequency responses

Time and frequency are related. If one matches, the other will too.
