Digital Control of a DC Servo Motor

ECE 461/661 Controls Systems Jake Glower - Lecture #34

Please visit Bison Academy for corresponding lecture notes, homework sets, and solutions

Problem:

- Control the speed of a DC servo motor
- Control the position of a DC servo motor

The mathematical model from before:

• Clifton 000-053479-002

$$\boldsymbol{\omega} \approx \left(\frac{39.28}{s+6}\right) V_a$$
$$\boldsymbol{\theta} \approx \left(\frac{39.28}{s(s+6)}\right) V_a$$

With T = 20ms

$$\boldsymbol{\omega} \approx \left(\frac{0.7404}{z - 0.8869}\right) V_a$$
$$\boldsymbol{\theta} \approx \left(\frac{0.02z}{z - 1}\right) \left(\frac{0.7404}{z - 0.8869}\right) V_a$$



Speed (rad/sec)

Hardware Setup

- Use a microcontroller to compute speed (or angle),
- Also use it to implement K(z)



Speed Control: Gain Compensation

$$G(s) = \left(\frac{39.28}{s+6}\right)$$

T = 20ms $G(z) = \left(\frac{0.7404}{z - 0.8869}\right)$ K(s) = k $GK = \left(\frac{0.7404k}{z - 0.8869}\right)$



Predicted Response

$$(GK)_s = -1$$

•
$$k = 0.3875$$

$$z = 0.2$$

•
$$k = 0.9277$$



Experimental Results:

z =0.6 • k = 0.3875z = 0.4• k = 0.6576z = 0.2• k = 0.9277Code: while(1) { : E = REF - SPEED;U = 0.9722 * E;D2A(U); Wait_20ms(); ļ



Speed Control: I Compensation

$$K(z) = \left(\frac{kz}{z-1}\right)$$
$$GK = \left(\frac{0.7404k \cdot z}{(z-1)(z-0.8869)}\right)$$



Type-1 System

• No Error for a Step Input

Code:

```
while(1) {
    :
    :
    E = REF - SPEED;
    U = U + k * E;
    D2A(U);
    Wait_20ms();
    }
```

I Compensation

$$z = 0.99$$
• $\left(\frac{0.7404k \cdot z}{(z-1)(z-0.8869)}\right)_{z=0.99} = -1$
• k = 0.0013

$$z = 0.95$$

• $\left(\frac{0.7404k \cdot z}{(z-1)(z-0.8869)}\right)_{z=0.95} = -1$

• k = 0.0045

$$z = 0.9434 + j0.0535$$

• $\left(\frac{0.7404k \cdot z}{(z-1)(z-0.8869)}\right)_z = -1$

•
$$k = 0.0082$$



Experimental Results

• I Compensation

Steady-state speed = 10 rad/sec

- No Error for a step input
- Type-1 system

The response is what the root locus plot predicted



Speed Control: PI Control

$$K(z) = k \left(\frac{z - 0.8869}{z - 1} \right)$$
$$GK = \left(\frac{0.7404k}{z - 1} \right)$$

Code:

```
while(1) {
    :
    E1 = E0;
    E0 = REF - SPEED;
    U = U + k * (E0 - 0.8869*E1);
    D2A(U);
    Wait_20ms();
    }
```



PI Control: Experimental Results

z = 0.99 k = 0.0135 z = 0.9 k = 0.1351 z = 0.6k = 0.5402

Again, the results are what the root locus plot predicts



Position Control:

• Change the sensor to an angle sensor and you have position control

$$\Theta = \left(\frac{39.28}{s(s+6)}\right) V_a = \left(\frac{0.0148z}{(z-1)(z-0.8869)}\right) V_a$$



Position Control: K(z) = k

$$\theta = \left(\frac{0.0148z}{(z-1)(z-0.8869)}\right) V_a$$

$$z = 0.99$$

$$k = 0.2291$$

$$z = 0.95$$

$$k = 0.458$$

$$z = 0.95 + j0.05$$

$$k = 1.1456$$



Position Control: K(s) = k

- Experimental Results
- In theory, the steady-state error is zero
 - Type-1 System

In practice, static friction causes a slight error

Otherwise, the response is what the root locus plot predicts



Lead Compensation

- Cancel the pole at s = -6
- Replace it with a pole at s = -19.2

$$K(z) = k\left(\frac{z - 0.8869}{z - 0.3}\right)$$
$$GK = \left(\frac{0.0148z}{(z - 1)(z - 0.3)}\right)$$

z = 0.4514 + j0.3102k = 26.8 $K(z) = 26.8 \left(\frac{z - 0.8869}{z - 0.3}\right)$



Lead Compensation:

$$K(z) = 26.8 \left(\frac{z - 0.8869}{z - 0.3}\right)$$

Again, the experimental results are what root locus plots predicted

- Slight steady-state error due to static friction
- (nonlinear terms)

Code:

```
while(1) {
    :
    E1 = E0;
    E0 = REF - SPEED;
    U = 0.3*U + k * (E0 - 0.8869*E1);
    D2A(U);
    Wait_20ms();
    }
```



Summary

Root locus really works

- It predicts how the system will behave as the gain changes
- The response is as the root locus plot predicts

Digital Control also really works

- It saves hardware: you don't need to build an op-amp circuit
- It removes the DC offset that op-amps have
- Download a new program and you have a new controller